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Manifold-Mapping Optimization Applied to Linear Actuator Design

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Optimization procedures, in practice, are based on highly accurate models that typically have an excessive computational cost. By exploiting auxiliary models that are less accurate, but much cheaper to compute, space mapping (SM) has been reported to accelerate such procedures. However, the SM solution does not always coincide with the accurate model optimum. We introduce manifold mapping, an improved version of SM that finds this precise solution with the same computational efficiency. By two examples in linear actuator design, we show that our technique delivers a significant speedup compared to other optimization schemes.

Index Terms—Actuators, optimization methods, space mapping (SM), surrogate-based optimization.

I. INTRODUCTION

Many practical design problems in electromagnetics are nowadays solved by means of highly accurate computational models. Typically, finite elements (FEs) are employed. However, this approach is computationally expensive. The space-mapping (SM) technique [1] aims at accelerating such procedures using approximate models that are faster to compute. The application of SM to optimization in low-frequency electromagnetics was considered, e.g., in [2]. The originally conceived SM technique might fail to converge to the accurate optimum. We introduce manifold mapping (MM) [3], an improved SM technique that efficiently finds the desired optimum. Its application is illustrated with some examples.

II. SM AND MM

A. SM Technique

Let us consider an optimization problem in the design space $X \subset \mathbb{R}^n$ with specifications $y \in Y \subset \mathbb{R}^m$. The accurate behavior of electromechanical devices is often studied using models that have large computational costs, e.g., FE models. In SM terminology, these models are called fine models. The fine model response is denoted by $f(x)$, where $x \in X$ is the design variable. The associated optimum is referred to as $x_f^*$. SM needs a second, possibly less accurate but computationally much cheaper model. This is the coarse model; in this paper, the coarse models are assumed to be defined over the same design space $X$. The coarse model response and the corresponding optimum are denoted by $c(x)$ and $x_c^* \in X$, respectively.

The SM function $p: X \rightarrow X$ is defined by

$$p(x) = \arg\min_{x \in X} ||c(p(x)) - f(x)||.$$  \hspace{1cm} (1)

SM corrects $x_c^*$, the coarse optimum, in order to approximate $x_f^*$. In traditional SM, this is done either by solving the nonlinear system of equations

$$p(x_{sm}^*) = x_c^*.$$  \hspace{1cm} (2)

or, more generally, by minimizing with the surrogate $c(p(x))$

$$x_{sm2}^* = \arg\min_{x \in X} ||c(p(x)) - y||.$$ \hspace{1cm} (3)

However, the solution of neither of these two problems does necessarily coincide with the fine model optimum $x_f^*$.

B. MM Technique

Analyzing the conditions under which the SM solution yields the correct answer, the SM technique can be improved in such a way that the accurate optimal design can be computed by an iterative process. Each step in this scheme requires one evaluation of the fine model $f(x)$ and one solution of a cheap minimization problem.

MM replaces the SM function $p(x)$ by an arbitrary bijection $\tilde{p}(x)$ and introduces the manifold mapping $S: c(X) \rightarrow f(X)$. With this mapping, the point $c(\tilde{p}(x_f^*))$ is mapped to $f(x_f^*)$ and the tangent space for $c(x_f^*)$ at $\tilde{p}(x_f^*)$ to the tangent space for $f(x_f^*)$ at $x_f^*$. By this construction, then

$$x_f^* = \arg\min_{x \in X} ||S(c(\tilde{p}(x))) - y||.$$ \hspace{1cm} (4)

In the examples in this work, the function $\tilde{p}(x)$ is taken equal to the identity.

C. Implementation of MM

During the iterative solution process, the manifold mapping $S$ is approximated by a sequence of affine mappings $S_k$, $k \geq 0$. This leads to the following algorithm (with $\tilde{p}: X \rightarrow X$ an arbitrary bijection).

1) Set $k = 0$ and compute

$$x_0 = \arg\min_{x \in X} ||c(\tilde{p}(x)) - y||.$$ \hspace{1cm} (5)

2) Evaluate $f(x_k)$ and $c(\tilde{p}(x_k))$ and finish if appropriate stopping criteria are met.

3) If $k > 0$ and with $\Delta c_k = c(\tilde{p}(x_{k-1})) - c(\tilde{p}(x_k))$ and $\Delta f_k = f(x_{k-1}) - f(x_k)$, $i = 1, \ldots, \min(n, k)$, we define $\Delta C$ and $\Delta F$ to be the $m \times \min(n, k)$ matrices with, respectively, $\Delta c_i$ and $\Delta f_i$ as columns. Their singular value decompositions are, respectively, $\Delta C = U_i \Sigma_i V_i^T$ and $\Delta F = U_f \Sigma_f V_f^T$.  

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4) The approximate affine mapping is

\[ S_k \mathbf{v} = f(\mathbf{x}_k) + S_k(\mathbf{v} - c(\mathbf{p}(\mathbf{x}_k))) \quad \forall \mathbf{v} \in Y \]  

where \( S_k \) is a regular \( m \times m \) matrix equal to the identity for \( k = 0 \) and to \( \Delta F \Delta C^T + (I - U_f U_f^T)(I - U_c U_c^T) \) if \( k > 0 \). The operator \( \dagger \) denotes the pseudoinverse.

5) Compute

\[ \mathbf{x}_{k+1} = \arg \min_{\mathbf{x} \in X} \| S_k(\mathbf{p}(\mathbf{x})) - \mathbf{y} \|. \]

6) Set \( k = k + 1 \) and go to 2.

Step 5) can be implemented in a simpler way since it can be shown that it is asymptotically equivalent to the following:

5) Compute

\[ \mathbf{x}_{k+1} = \arg \min_{\mathbf{x} \in X} \| c(\mathbf{p}(\mathbf{x})) - \tilde{\mathbf{y}}_k \| \]

where

\[ \tilde{\mathbf{y}}_k = c(\mathbf{p}(\mathbf{x}_k)) - (\Delta C \Delta F^T + I - U_c U_c^T)(f(\mathbf{x}_k) - \mathbf{y}). \]

The matrix \( S_k \) satisfies

\[ S_k(c(\mathbf{p}(\mathbf{x}_k)) - c(\mathbf{p}(\mathbf{x}_k))) = f(\mathbf{x}_k) - f(\mathbf{x}_k) \]

for \( k > 0 \) and \( I = \max(0, k - n), \cdots, k - 1 \), and, thus, in the limit, the desirable properties of the manifold mapping \( S \) are fulfilled. Under convergence of the scheme given, the limit is the fine model optimum \( \mathbf{x}^* \).

III. AUTOMOTIVE MAGNETIC ACTUATOR

For this first actuator design problem, SM, MM, and some other efficient optimization techniques are compared. Automotive actuators typically generate high levels of force and can be used in devices, such as electromagnetic switches, relays, valves, etc. Fig. 1 is a schematic view of the cylindrical plunger electromagnet. It was originally introduced in [4], and it consists of a core and a plunger, both made of iron, and a copper coil. The sizes \( \mathbf{x} = (x_1, x_2) \) have to be optimized in order to assure a maximum magnetic thrust force versus displacement in the \( z \) axis. The volume of the device is kept constant. The specifications \( \mathbf{y} \) are a constant force response of 100 \( N \), for six vertical plunger displacements. The design space \( X \) proposed in [4] is the rectangle \( [2.5, 4.5] \times [12, 18] \), where all the bounds have been specified in millimeters. More details on the problem can be found in [4, 5].

The fine model \( f(\mathbf{x}) \) is based on a static force computation for the six plunger displacements. This force is evaluated numerically by second-order Lagrangian FEs [6] with one level of adaptive refinement. The number of degrees of freedom in every case is around 5000–7000 yielding three digits of accuracy in the force. The objective function here is \( F(\mathbf{x}) = \| f(\mathbf{x}) - \mathbf{y} \|_2 \), i.e., a relative 2–norm of the model discrepancy.

The coarse model \( c(\mathbf{x}) \) is based on a magnetic equivalent circuit (MEC) [7] derived from the actuator. The MEC is nonlinear, in the sense that the \( B-H \) curve is considered for the plunger. The core magnetic permeability is taken as infinite and neither fringing nor leakage effects are included. A linear interpolation for the force is taken between the points for maximum and minimum displacement. This coarse model is around 400 times faster than the fine one. In Fig. 2, the fine and coarse model responses are shown for two points in the design space \( X \): the geometrical center \( \mathbf{x}_0 \) and the optimum obtained with MM. Each optimization involving the coarse model is solved by sequential quadratic programming (SQP) [8].

Table I shows the performance of seven minimization methods when applied to the mentioned fine model. The given figures (amount of work expressed in the number of equivalent fine model evaluations) are approximately proportional to the total computing time. All seven techniques yield solutions with similar objective function values. The first two (MMA: method
of moving asymptotes and PM: penalty method) correspond to methods analyzed in [4] and the next two (EM: ellipsoid method and RQP: recursive quadratic programming) to other schemes also studied by the same author [5]. The coarse model optimum \( x_c^* = [4.5 \text{ mm}, 18 \text{ mm}] \) has a lower fine cost function than any of the solutions given by the methods above. SM and MM converge to \( x_c^* \) after just one fine model evaluation (the number of coarse model evaluations is also taken into account in the amount of work shown in Table I). SQP, applied to the fine model and starting from the coarse model optimum, performs also very efficiently.

Next, the space \( X \) is slightly modified to \([2.5,5] \times [15,30]\) (again, in millimeters) in order to have a design in which the coarse optimum does not end the optimization process.\(^1\) Thus, a better comparison can be made between the three methods that perform similarly above. Now, the coarse optimum is the point \([4.68 \text{ mm}, 17.72 \text{ mm}]\) which has an associated cost function of 465. The optimization results are given in Table II. MM and SQP yield both a solution with the same quality, but MM is almost four times faster. SM improves also SQP in efficiency but the result obtained is not the optimal design. This experiment corroborates MM as a method as efficient as SM, but with the additional property of convergence toward the fine optimum.

<table>
<thead>
<tr>
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<th># f evals.</th>
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<tbody>
<tr>
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<td>45.0</td>
<td>[5.00, 16.45]</td>
</tr>
<tr>
<td>MM</td>
<td>3.2</td>
<td>45.0</td>
<td>[5.00, 16.61]</td>
</tr>
<tr>
<td>SM</td>
<td>6.9</td>
<td>45.3</td>
<td>[5.00, 15.00]</td>
</tr>
</tbody>
</table>

### IV. CYLINDRICAL VOICE COIL ACTUATOR

#### A. Actuator Description

Linear voice coil actuators [9] are noncommutated electromechanical devices that provide cogging-free force outputs directly proportional to the applied current. An axisymmetrical variant consisting of a permanent magnet, a coil, and a ferromagnetic core is presented in Fig. 1. As in the previous section, the design variables \( \mathbf{x} \) correspond to sizes in the actuator. The distance \( D \) measures the position of the coil. For this actuator, we consider two design problems. Full details on both problems can be found in [10].

#### B. Force Response Optimization

In the first problem, we allow the coil to move over a 4-mm stroke, i.e., \( 0 \leq D \leq 4 \text{ mm} \). Values for \( \mathbf{x} \) have to be found such that the force response is as flat and as close to \( \mathbf{y} = 24 \text{ N} \) as possible. For this force level, the nonlinear effect of the ferromagnetic core has to be taken into account in an accurate fine model description. The permanent magnet demagnetization curve is assumed to be linear with \( \mu_r = 1 \). The coil movement is represented at nine equidistant points in the interval \( 0 \leq D \leq 4 \text{ mm} \). The cost function is \( F(\mathbf{x}) = ||f(\mathbf{x}) - \mathbf{y}||_2/||\mathbf{y}||_2 \times 100. \)

\(^1\)This modification of the design space is introduced for testing purposes. As observed in [5], big values for \( x_3 \) could lead to devices with significant mass and a poor dynamic regime performance.

In the fine model, the force is computed by Lorentz’s formula at each of the nine coil positions. Second-order Lagrangian triangular FEs are used for that purpose. The number of degrees of freedom is between 8000 and 11 000, yielding three digits of accuracy in the force.

The first of the two coarse models considered is an FE model in which the nonlinear material characteristic of the core is linearized. Depending on the number of Newton iterations required, this model is a factor between 30 and 50 cheaper than the fine one. The second coarse model is an MEC. This model has negligible computational cost compared to the fine model.

We initially consider the one-dimensional (1-D) design problem, considering changes in the first design variable \( x_3 \) only. The purpose is to illustrate how the cost function associated with MM closely approximates the fine model cost function in a region close to \( x_3^* \).

In this 1-D problem, both SM and MM, using either the linear FE or the MEC as coarse model, converge in four fine model evaluations. Compared with SQP or the Nelder–Mead Simplex (NMS) method [11], that are similar in performance, SM and MM deliver a speedup of a factor between four and five.

If we consider the two-dimensional (2-D) design problem, with changes in \( x_3 \) and \( x_2 \) only, we can clearly illustrate that the problem is ill conditioned. In Fig. 4, the cost function is plotted. In this plot, the dark region shows a long and steep valley in the design space with approximately the same cost function value. This means that there is no unique solution. The value found by optimization depends on factors such as starting guess and algorithm used. Uniqueness can be restored by regularization, for example, by imposing a minimal mass constraint as will be done in the following subsection.

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**TABLE II**

<table>
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**Fig. 3.** Cost functions associated with the coarse (simplified FE), fine, SM, and MM surrogate models for the 1-D problem.
provided that the constraints were met with three digits of accuracy. SM and MM show a similar behavior: convergence is reached in seven and six fine constraints evaluations, respectively. SQP converged within 56 fine constraint evaluations. Because, in this problem, the number of constraints is smaller than the number of design parameters, we expect the SM solution to coincide with $\mathbf{x}^*$. We recognize this to be the case as SQP, SM, and MM yield the same optimum, but even in this situation, MM offers an additional advantage over SM. The computation of the SM function $p(x)$ is a very delicate issue [1]. MM replaces it by the identity, and in a realistic problem like this one, the efficiency results are comparable.

V. CONCLUSION

We introduced the manifold-mapping technique, an optimization method that exploits approximate model information to accelerate time-consuming design procedures. The application of the method on linear actuators has given evidence of the computational speedup that this method can deliver.

ACKNOWLEDGMENT

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REFERENCES