An Industrial Streamer Corona Plasma System for Gas Cleaning

G. J. J. Winands, Keping Yan, A. J. M. Pemen, Member, IEEE, S. A. Nair, Zhen Liu, and E. J. M. van Heesch

Abstract—For pulsed corona plasma applications, it becomes important to develop pilot systems with large average power and high-energy conversion efficiency. Since the beginning of 2000, we have been working on an industrial corona plasma system with tasks of 10–30 kW in average power and higher than 90% of total energy conversion efficiency. The pulsed-power source should have the following specifications: rise time of 10–25 ns, pulsewidth of 50–150 ns, pulse repetition rate of up to 1000 pulses per second, peak voltage pulse of 70 kV, peak current of 3.5 kA, dc bias voltage of 10–35 kV, and energy per pulse of up to 30 J. Sixteen parallel wire cylinder reactors are used to match the source. Gas and reactor temperatures can be controlled individually with water flow around the outside of those cylinders. The system is designed for gaseous oxidation and electrostatic dust precipitation. The system has been used for up to 17 kW in average power. This paper reports the system in detail, discusses issues related to the matching between the source and the reactor, and presents an example of industrial demonstrations on odor abatement at 1000 m³/h. Finally, this paper also gives a general guideline for design of corona plasma systems.

Index Terms—Corona, impedance matching, spark gaps, transmission lines.

I. INTRODUCTION

A N INTERESTING technique for removal of multiple pollutants, such as volatile organic compounds (VOCs), odor, bacteria, tars, fine particles, NOₓ, and SOₓ, is to use streamer corona plasma [1]–[12]. Radicals produced by the discharge induce chemical reactions leading to a decrease in the pollutant concentrations. The pollutants are often converted to by-products, which are either less harmful or easy to be handled. It has been confirmed that the technique is very effective for low concentration abatement, its investment and operational costs are lower and multiple pollutants can be removed simultaneously.

The first industrial corona plasma system was reported by Ente Nazionale Energia Elettrica (ENEL) for simultaneous removal of dusts, SO₂, NOₓ, and heavy metals from exhaust gases [1]. Unfortunately, lack of cost effective corona plasma generation and processing techniques discouraged industries. Nevertheless, three industrial corona plasma demonstration systems with up to 40–120 kW in average power were recently reported in Japan, Korea, and China. For example the system described by Mok et al. [13], with an average output power of 40 kW, or the system described by Lee et al. [14], with an average output power of 120 kW. All of them are based on magnetic compression techniques with pulse duration of 200–500 ns. Main drawbacks of these reported systems are their relatively low-energy conversion efficiency.

Our previous pulsed corona plasma system gives an average output power of 2 kW with a pulsewidth of 20–100 ns [15], which is too small for industrial applications. The main objective of this paper is to report a novel pulsed corona plasma system with 10–30 kW of average power, 50–100 ns of pulse duration, and 90% of energy conversion efficiency.

II. PULSED-POWER SOURCE

For pulsed corona system, nonthermal plasma is produced by applying short high-voltage pulses onto a nonuniform electrode geometry. When the applied electric field in between the electrodes exceeds a critical value, streamers start propagating from one electrode toward the counter one. The energetic electrons (∼10 eV) around the streamer head dissociate gaseous molecules, generate radicals, and then induce chemical reactions. As soon as the primary streamer bridges the gap a secondary streamer is formed, which travels only partly into the gap, and is much less energetic. As a result, fewer radicals are produced and its chemical efficiency is lower. For efficient pollutant removal, it can be stated that primary streamers are more efficient than the secondary streamers, and as soon as the primary streamers cross the electrode gap, the applied electric field should be removed. Besides this argument, several other issues also have to be considered when constructing an industrial pulsed corona system, such as output power, lifetime, reliability, safety, and energy efficiency. Generally speaking, the issue of system processing efficiency can be divided into four parts as illustrated in Fig. 1.

The first step is related to the source design. The energy from the mains has to be modulated into pulses efficiently. The second step is about the matching between the source and the reactor. The modulated pulses have to be applied efficiently to the reactor. The third one is related to plasma generation. Transferring a large amount of energy into the reactor does not necessarily imply high chemical yield. The electron energy distribution function of the generated plasma will greatly affect the processing efficiency. The homogeneity of the generated
plasma is another important factor to influence the efficiency. The fourth step is mainly dependent on the related process. By changing process parameters, such as temperature and humidity, or by using additives and catalysts, one may optimize the induced reactions.

This paper does not intend to discuss all of the four steps in details, but to focus on the first two steps. The energy conversion efficiency of the first step is defined as the efficiency to charge the high-voltage pulse forming capacitor. It is calculated by the ratio of the energy stored inside the capacitor over the total energy input from the mains. For the present design, it is about 95% [4]. The second step is the efficiency from the modulator to the reactor, i.e., the ratio of corona energy over the stored energy. For achieving high efficiency, the output impedance of the pulse generator and the impedance of the reactor have to be properly matched. The problem here is caused by the impedance of the reactor, which is transient and nonlinear. The total energization process has to be considered, namely before, during and after streamer generation [4].

Before the applied voltage reaches the streamer inception, the reactor behaves as a capacitor with a vacuum capacitance of $C_r$. For a power modulator generating fast rising (rise time $\tau$) pulses, the rise time on the reactor becomes $2Z_{\text{out}} \cdot C_r$, where $Z_{\text{out}}$ is the output impedance of the modulator. Simulations showed that optimal energy conversion efficiency can be achieved when the plasma is started before the moment at which the energy transfer from the source to this capacitance reaches its maximum value. The exact timing depends on the values of $C_r$, $Z_{\text{out}}$, and $\tau$. Igniting the plasma too late will result in low-energy conversion efficiency. During streamer generation, the reactor can be depicted as a resistor $R_t$. When $R_t = Z_{\text{out}}$, the energy efficiency reaches its maximum. It has been observed that the impedance $R_t$ tends to decrease to the output impedance $Z_{\text{out}}$ of the generator when increasing the peak voltage on the reactor. Superposition of a dc bias voltage on the voltage pulses can also be used to increase the voltage, and thus to improve the matching. After corona plasma quenching, the current rapidly drops to zero, and the reactor impedance becomes much larger than the output impedance $Z_{\text{out}}$. As a result, the energy transfer stops and a high-frequency oscillation can be observed. To prevent damage of the switch as a result of this oscillation, a transmission line transformer (TLT) with two-type magnetic cores is used to absorb the reflected energy.

### A. Hybrid Pulsed-Power System (HPPS) System

Fig. 2 shows a schematic representation of the HPPS. The term hybrid was used because not only pulses are applied onto the reactor but also a dc bias voltage is used. The system is based on resonant charging of a capacitor bank $C_h$ and subsequently discharging it via a heavy-duty spark-gap switch. The dc source not only results in an increase of the energy per pulse and an improvement of the energy efficiency, but also enables electrostatic precipitation (ESP). The general characteristics of the system are summarized in Table I.

<table>
<thead>
<tr>
<th>TABLE I</th>
<th>GENERAL CHARACTERISTICS OF THE HPPS SYSTEM</th>
</tr>
</thead>
<tbody>
<tr>
<td>Average Power</td>
<td>10 - 30 kW</td>
</tr>
<tr>
<td>Rise time</td>
<td>15 ns</td>
</tr>
<tr>
<td>Max. pulse voltage</td>
<td>70 kV</td>
</tr>
<tr>
<td>Pulse width (power)</td>
<td>50 - 150 ns</td>
</tr>
<tr>
<td>Peak current</td>
<td>3.5 kA</td>
</tr>
<tr>
<td>Energy per pulse (EPP)</td>
<td>10 - 30 J</td>
</tr>
<tr>
<td>Pulse repetition rate</td>
<td>1-1000 pps</td>
</tr>
<tr>
<td>Jitter on EPP</td>
<td>1~2%</td>
</tr>
<tr>
<td>DC bias voltage</td>
<td>0-35 kV</td>
</tr>
<tr>
<td>Energy efficiency</td>
<td>~90%</td>
</tr>
</tbody>
</table>

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The coupling capacitor \( \mu \) and current pulses of the low-voltage part are around 25 \( \mu \)s with a jitter of less than 1% [4]. The pulsewidth, the output voltage \( C \) and \( \text{LCR} \) transformer ratio.

1) Voltage Pulse Generator: With regard to the circuit in Fig. 2: the coupling capacitor \( C_{\text{DC}} \) separates the circuit into two parts. The left side is the voltage pulse modulator, and the right side is the dc source. The low-voltage part consists of a mains filter, a set of rectifiers, three air-core inductors (10–30 \( \mu \)H), three thyristors, and two energy storage capacitors (\( C_0 = 2–3 \) mF). The three thyristors are switched consecutively in order to resonantly charge \( C_h \). The third thyristor is used to stabilize the voltage on \( C_L \), which results in an output with a jitter of less than 1% [4]. The pulsewidth, the output voltage and current pulses of the low-voltage part are around 25 \( \mu \)s, 1 kV, and 1 kA, respectively. A step-up pulse transformer is used to increase the voltage pulse to 36 kV. For proper matching between the low- and the high-voltage parts, the values of \( C_L \) and \( C_h \) obey the equation of \( C_L = N^2C_h \), where \( N = 36 \) is the transformer ratio.

The most critical component of the voltage pulse generator is the heavy-duty switch used to discharge the high-voltage capacitor \( C_h \) into the TLT. The switch should satisfy the following requirements: high hold-off voltage, nanosecond switching time, an impedance that matches the input of the TLT, high switching current, high pulse repetition rate, low losses, long lifetime, cost effective, compact, and insensitive to occasional over voltage/current. In our opinion, advanced spark-gap switches do match those specifications. For the present HPPS system, a high-pressure spark-gap switch with switching voltage of 40 kV, switching current of 4 kA, and pulse repetition rate of 1000 pps is used. Its main specifications are illustrated in Fig. 3. The switch can be pressurized up to 4 bars, and is flushed with dry air of up to 35 N \cdot m^3/h to improve the recovery time [16]. Copper-tungsten alloy is used as electrode material. The estimated lifetime is in the range of \( 10^9–10^{10} \) shots at 10 J/pulse. It was constructed coaxially to make its impedance equal to the input impedance of the TLT. Such optimal design creates short arcs, which dissipate less energy and have lower inductance [17].

For stable operation, the switch should close just after the high-voltage capacitor is charged. Prefiring decreases the average switching voltage, and thus causes a decrease of the energy. Late firing implies long-period stress on the high-voltage components, and limits the pulse repetition rate. With regard to the circuit in Fig. 2, the triggering is performed with an \( LCR \) trigger method, where a 10-\( \mu \)H inductor, a 450-pF capacitor, and a 1.2-M\( \Omega \) resistor form the trigger circuit. Details were reported earlier [4], [18].

After the switch closes, the capacitor \( C_h \) is discharged into a two stage TLT, which has an input and output impedance of \( Z_0/2 \) and \( 2Z_0 \), respectively. The TLT was constructed from concentric metal tubes with \( Z_0 \sim 10 \) \( \Omega \), and it is filled with transformer oil. For large average output power and short pulsewidths, it is preferable to keep the output impedance of the source \( (Z_{\text{out}}) \) as low as possible. For a capacitor discharge, the pulsewidth is roughly equal to \( Z_{\text{out}} \cdot C_h \). For high average power, \( C_h \) has to be large, thus \( Z_{\text{out}} \) should be small. A TLT usually consists of a set of transmission lines connected in series or in parallel configurations at its ends. It can be used for several purposes, such as to match impedances and to multiply voltage and/or current. Very detailed discussions can be found in [19]. For the present system, it is used to increase the output voltage. The ideal gain of an \( n \)-stage TLT is equal to its line number \( n \). In practice, the gain is a little smaller. To minimize the losses, the secondary mode impedance \( Z_s \), which is defined as the characteristic impedance between the outer conductor of the upper cable and the ground structure, has to be as large as possible. The corresponding current is called secondary mode current. The secondary mode current has to be as small as...
possible in order to achieve an optimal gain and high-energy efficiency. In practice, this is often accomplished by placing magnetic cores around the upper stages of the TLT.

2) DC Bias Power Source: A 1–2–mH ferrite core inductor \( L_{dc} \) is used to decouple the voltage pulse from the dc. In order to avoid spark breakdown, the dc base voltage is limited to 30 kV for the present system. When using the dc source, two sources will energize the reactor independently. Before a high-voltage pulse is applied, the reactor is charged to \( V_{dc} \) by the dc source in about 100 \( \mu s \) via a charging diode \( D_{dc} \), inductor \( L_{dc} \), coupling capacitor \( C_{dc} \) (14 nF), and the TLT. When the spark gap is fired, the energy stored in both \( C_{dc} \) and \( C_h \) is transferred to the reactor simultaneously. After plasma generation, the reactor is charged again by the dc source. During the pulsed energization, streamer corona plasma is generated; and in between pulses ESP can be realized.

**B. Streamer Corona Plasma Reactor**

For industrial applications, wire-cylinder, wire-plate, or multipoint-plate configurations are commonly used. The electrode gap distance and electrode(s) curvatures are determined by the output voltage of the source. Ideally, the streamers should cross the electrode gap completely. This requires a minimum voltage and pulsewidth. For industrial applications, the electrode gap distance is preferred to be large. The distance, however, is limited because pulsed-power sources with extreme high-voltage output are expensive. The length of the high-voltage electrode depends on the required energy and on the type of processes. For nontermination processes [4] radicals mainly react with pollutants. In this case, the removal efficiency linearly increases with applied energy, and the process does not depend on the residence time. A smaller reactor with higher energy density per pulse per unit length of the reactor can be adopted. For nonlinear termination processes, however, radicals mainly terminate due to reactions between themselves. In this case, it is preferable to increase the reactor volume and to reduce the energy density per pulse and per unit volume of the reactor. This is also true for particle collection. Specific guidelines for designing the reactor can be found in [24].

The reactor of the HPPS consists of 16, vertically positioned, wire-cylinder reactors in parallel. The wire diameter is 3 mm. The length and diameter of each reactor are 1000 and 160 mm, respectively. This results in a total reactor volume of 0.32 m\(^3\). The reactor is constructed from stainless steel to prevent degradation as a result of chemical attack from the created acids. The outside of the cylinders can be cooled with water. Gaseous inlet and outlet are located at the bottom and the upper of the reactor, respectively. The reactor is energized from its top with a maximum energy per pulse per meter of 2 J.

**III. RESULTS AND DISCUSSIONS**

To evaluate the voltage and the current waveforms at the input of the reactor, the experimental set-up is equipped with differentiating–integrating (D–I) measuring systems [20]. A digital oscilloscope (1 GHz, 4 GS/s) is used to record the measured signals. The power waveform is determined as the product of the voltage and current waveforms. The energy is calculated by integrating the power waveform. Fig. 4 shows the results.

The rise time of the pulse on the reactor, the time required for the voltage to rise from 10% to 90% of its maximal value, is about 16 ns (equivalent 1/e-time is 12 ns). As expected, the rise time does not depend on the energy per pulse, dc bias voltage and pulse repetition rate since the rise time (1/e-time) of the voltage on the reactor is equal to \( 2Z_{out}C_r \) (assuming that the pulses generated by the source have rise-times \( \ll 2Z_{out}C_r \)). The reactor capacitance \( C_r \) was determined to be 320 pF. Combined with the output impedance \( Z_{out} \) of 20 \( \Omega \) this gives a rise time of 13 ns which is close to the experimental value. The agreement between experiment and theory indicates that the rise time of the pulses generated by the power modulator is much smaller than the characteristic rise time of the voltage on the reactor load. The pulsewidth [full-width at half-maximum (FWHM)] is around 50–60 ns. It was observed that the pulsewidth depends on the energy per pulse, as can be expected for a capacitor discharge.

In Fig. 5(a), it can be seen that the peak power on the reactor increases as a function of the dc bias \( V_{dc} \). Because the energy per pulse is the time integral of the power waveform, it is obvious that the energy per pulse also increases. This increase is due to the following two factors. First, as expressed by (1), there is more stored energy \( E_{dc} \) in the coupling capacitor \( C_{dc} \) and in the reactor:

\[
E_{dc} = \frac{1}{2}(C_{dc} + C_r)V_{dc}^2. \tag{1}
\]

Second, by increasing the dc bias voltage, more of the stored energy in \( C_h \) can be transferred to the reactor as a result of improved matching. This phenomenon can be observed by
studying the voltage waveforms on $C_h$ as shown in Fig. 5(b) and (c). The capacitor is resonantly charged in about 50 µs to its maximum $V_{\text{max}}$. The voltage, then, remains almost constant until the spark-gap switch is fired. Some ringing occurs after the switch is closed due to the mismatch. Fig. 5(c) replots the waveforms on a short time scale. One can easily see that the capacitor does not discharge completely, but to a residual voltage $V_{\text{end}}$, indicating that not all of the stored energy can be transferred into the reactor within the short duration. Thus, the energy from $C_h$ used for the plasma generation becomes

$$E_{C_h} = \frac{1}{2} C_h \left( V_{\text{max}}^2 - V_{\text{end}}^2 \right).$$  

(2)

By superimposing a dc bias voltage, both the residual voltage $V_{\text{end}}$ and its ringing decrease, and thus much more energy is transferred to the reactor. Obviously, the matching is improved. Besides the economical advantage of having more energy going from $C_h$ into the reactor, there is also the advantage of decreased component wear. Most of the energy that does not go into the reactor is dissipated in the spark-gap switch, the TLT and in the $D_2$–$R_5$ circuit. An example of the energy dissipation in $R_5$ is shown in Fig. 5(d). We also observed that either increasing the dc bias voltage or pulse voltage amplitude can lead to an improvement in matching. A typical example is shown in Fig. 5(e) [4]. Clearly, for sufficiently high voltage, the reactor impedance becomes equal to the output impedance of the source and the energy transfer can thus be optimized. It seems that the plasma generation is automatically adjusted to match the available energy stored in the capacitors. Changes in plasma generation as a function of the applied voltage have already been observed [21] using a fast intensified charge-coupled device (CCD) camera. It is believed that changes in streamer diameter, velocity, and branching lead to the automatic impedance matching.

The results shown in Fig. 6 summarize the experiments performed to optimize the output power and the energy efficiency. The experiments were carried out during 5-min test cycles. Long duration experiments were also performed. It was observed that the jitter in average output energy was about 1%–2%. No problems occurred during the long duration operation. The output power is varied by varying the pulse repetition rate, the $C_h$ value, and the dc-bias voltage. The results are summarized in Fig. 6(a) with a maximum average power of 17 kW. The energy per pulse decreases slightly as function of pulse repetition rate due to the small value of the primary energy storage capacitor $C_0$. 

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Fig. 5. (a) Power waveforms as function of the dc bias voltage, $C_h = 10.6$ nF. (b) Voltage waveform on $C_h$ (= 25.8 nF), single-shot measurement without dc bias voltage. (c) Voltage waveforms on $C_h$ after the spark-gap fires as function of dc bias voltage level, single-shot measurements. (d) Energy dissipation in $R_5$ resistor as function of the dc bias voltage, $C_h = 10.6$ nF. (e) The ratio of the reactor impedance over the output impedance as function of applied voltage. The reactor impedance is defined as the ratio of voltage over current at the moment the current reaches its maximum value. Two reactor configurations were used, both wire-cylinder type. Length, wire diameter and cylinder diameter are 3000, 3, and 160 mm for the circles, and 3000, 1, and 250 mm for the triangles, respectively. The solid line is an interpolation of the measuring results.
The energy efficiency (of the second step in the chain depicted in Fig. 1) is defined as the ratio of primary stored energy over the corona energy as

\[ \eta = \frac{\int V_{\text{reactor}} \cdot I_{\text{reactor}} \, dt}{\frac{1}{2} (C_{\text{DC}} + C_I) V_{\text{dc}}^2 + \frac{1}{2} C_h V_{\text{max}}^2}. \] (3)

It does not depend on the pulse repetition rate, but increases with dc bias voltage due to the improved matching. A maximum energy efficiency of about 95% was obtained. Considering a 95% of the efficiency from the mains to the \( C_h \), we can conclude that the total energy conversion efficiency is about 90%. For the 120 kW average power modulator described in [14], the efficiency is about 50%. The 40-kW average power modulator described in [13] has a efficiency of less then 76%; and for the system described in [22], it is less then 50% for 1–2 kW average power. All systems have considerably lower energy transfer efficiency than the presented one. Part of the difference between the efficiency of these systems is due to the use of magnetic pulse compression technique. The other difference is due to the system design and matching.

A. Industrial Demonstration on Odor Emission Abatement

As an example of industrial applications of the HPPS system, Table II lists results of an industrial demonstration on odor emission abatement for a Dutch compost manufacturer. The experiments were performed as a feasibility study to evaluate simultaneous removal of various gaseous compounds and the present system design. Fig. 7 shows a diagram of the related production process. An acid washer and a biobed filter were already used, but their performance did not match the emission regulation. The total flow rate of the exhaust gas is about 250,000 N · m³/h. Though our plasma system could be installed at three locations as indicated in Fig. 7, we mainly focused on its performance where odor concentrations are high, i.e., position one, just after the manufacturing process. The demonstration was performed with a substream gas flow of
1000 N·m³/h, and with a corona energy per pulse of 8 J, a pulse repetition rate of 250 pps, a peak-voltage on the reactor of 70 kV, a peak current of 3 kA, a pulsewidth (FWHM) of 75 ns, and a rise time of the voltage pulse of 16 ns. The total corona plasma power is 2 kW, resulting in a corona energy density of 7.2 J/L. The residence time of the gas inside the reactor was 1.2 s. The gas temperature was about 30 °C–40 °C, the relative humidity was higher than 80%. The demonstration lasted several hours continuously without any problem.

Chemical analyses were performed by a certified lab, PRA OdourNet BV. For the analyses, 40 L of gases from the output of the reactor were sampled in naphthene bags in 30-min experimental measuring cycles. The bag was placed inside a low-pressure container. The gas is sampled according to the “lung-method” to prevent the odorous gases to come into direct contact with the pump. To prevent condensation inside the bags and its related absorption, the sampled gas was diluted with nitrogen. Several techniques were used for gaseous analysis. The Impinger method was used to determine H₂S concentration. A gas chromatograph mass spectrometer (GCMS) was used to determine VOCs concentrations. The GCMS consists of a Varian 2700 gas chromatograph and a MAT 112S Finnigan mass spectrometer.

The removal efficiency of H₂S exceeded 95% (initial concentration 15.3 µg/m³, output concentration below detection limit of 0.8 µg/m³). Table II presents the results for VOC removal in terms of initial and final concentrations. One can easily see that multiple gases can be removed simultaneously. Aldehydes are the most difficult ones to be removed. Esters on the other hand, are produced by the plasma treatment. We also observed that the plasma leads to a certain amount of ozone slip when increasing the plasma power.

In comparison with Wang’s [23] work (H₂S removal under a gas flow rate of 5.7 m³/min, less than 76.0 µg/m³ of H₂S, a discharge power of 1.2 kVA, a removal efficiency of 95%, and a corona energy of 12.6 J/L), the present system has an energy consumption of about 57% of the reported one. Although pollutant removal efficiency is usually dependent on initial pollutant concentration, and as a consequence the obtained H₂S removal efficiency cannot be compared directly to the work of Wang, in general removal efficiency decreases for decreasing concentrations. The obtained decrease of energy consumption is thus a lower threshold value.

IV. Conclusion

DC bias superimposed pulse energization leads not only to an increase of the energy per pulse but also to an improvement of the matching between reactor and power source. Industrial plasma systems with 90% energy conversion efficiency can be designed by following the proposed guidelines. Industrial demonstration on odor emission abatement for a Dutch compost manufacturer proved the design and also confirmed the feasibility of simultaneous removal of multiple compounds.

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REFERENCES

Keping Yan was born in Kerkrade, The Netherlands, in 1978. He received the B.S. and M.S. degrees in applied physics from the Eindhoven University of Technology, The Netherlands, in 2002. In 2002, he joined the faculty of Electrical Engineering at the Eindhoven University of Technology. His Ph.D. work focuses on the interaction between power modulator and plasma generation/chemical processing. His interests are related to repetitive plasma generation for industrial scale gas-cleaning applications.

G. J. J. Winands was born in Kerkrade, The Netherlands, in 1978. He received the M.Sc. degree in applied physics from the Eindhoven University of Technology, The Netherlands, in 2002. In 2002, he joined the faculty of Electrical Engineering at the Eindhoven University of Technology. His Ph.D. work focuses on the interaction between power modulator and plasma generation/chemical processing. His interests are related to repetitive plasma generation for industrial scale gas-cleaning applications.

A. J. M. Pemen (M’98) received the B.Sc. degree from the College of Advanced Technology, Brede, The Netherlands, in 1986 and the Ph.D. degree in electrical engineering from the Eindhoven University of Technology, The Netherlands, in 2000, both in electrical engineering. Before joining the Electrical Power Systems Group (EPS), Eindhoven University of Technology in 1998 as an Assistant Professor, he was with KEMA T&D Power, Arnhem, The Netherlands. He is currently involved in research on pulsed-power and pulsed plasma. His research interest includes high-voltage engineering, pulsed-power, plasmas, and renewable energy systems. Among his achievements are the development of an on-line monitoring system for partial discharges in turbine generators, a pulsed-corona system for industrial applications, and a pulsed corona tar cracker. He is the founder of the Dutch Generator Expertise-Center.

S. A. Nair was born in India, in 1978. He received the B.S. in chemical engineering from the Institute of Chemical Technology, University of Mumbai, Mumbai, India, and the M.S. degree from the Eindhoven University of Technology, Eindhoven, The Netherlands, in 2000 and 2004, respectively. Since 2004, he has been a Japan Society for Promotion of Science Research Fellow with the Tokyo Institute of Technology, Tokyo, Japan. His research interests include understanding the chemical processes in nonthermal plasma-assisted catalytic fuel reforming (methane and biofuels) and nonthermal plasma-assisted gas cleaning, with a focus on developing catalysts, kinetic modeling, and design of reactors.

Zhen Liu was born in Xiangcheng, China, in 1978. He received the B.S. degree from the Department of Electronic Science and Technology, Xi’an Jiaotong University, China, in 2000 and the M.S. degree from the Electrical Department, Tsinghua University, China, in 2003. He is currently working toward the Ph.D. degree at the Electrical Power Systems Group (EPS), Technische Universiteit Eindhoven (TU/e), The Netherlands. His research interest is repetitive multiple-switch pulsed-power generation.

E. J. M. van Heesch was born in Utrecht, The Netherlands, in 1951. He received the Master’s degree in physics from the Eindhoven University of Technology, The Netherlands, and the Ph.D. degree in plasma physics and fusion related research from the University of Utrecht, The Netherlands, in 1975 and 1982, respectively. Since 1986, he has been an Assistant Professor at the Eindhoven University of Technology. Here, he is leading pulsed-power research. He was previously involved with shock-tube gas dynamics (Eindhoven, 1975) and with fusion technology (Jutphaas, The Netherlands, 1975–1984, Suchumi former USSR, 1978 and Saskatoon, Canada, 1984–1986). Among his design's are various plasma diagnostics, a toroidal fusion experiment, substation high-voltage measuring systems and systems for pulsed-power processing. He organizes many projects with industry and national and European Union research agencies. His research is the basis for teaching and coaching university students and Ph.D. candidates. He is a Co-inventor of several patents and has coauthored more than 100 publications.