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A Closed-Form Solution to Estimate Spatially Varying Parameters in Heat and Mass Transport

Ricky J. R. van Kampen^{ID}, Amritam Das^{ID}, Siep Weiland^{ID}, and Matthijs van Berkel^{ID}

Abstract—This letter presents a closed-form solution to estimate space-dependent transport parameters of a linear one dimensional diffusion-transport-reaction equation. The infinite dimensional problem is approximated by a finite dimensional model by 1) taking a frequency domain approach, 2) linear parameterization of the unknown parameters, and 3) using a semi-discretization. Assuming full state knowledge, the commonly used output error criterion is rewritten as the equation error criterion such that the problem results in linear least squares. The optimum is then given by a closed-form solution, avoiding computational expensive optimization methods. Functioning of the proposed method is illustrated by means of simulation.

Index Terms—Grey-box modeling, distributed parameter systems, identification, inverse problems.

I. INTRODUCTION

HEAT and mass transport phenomena are widely studied in the domain of physics and chemistry. Examples include, but are not limited to, transport of thermal energy through nuclear fusion reactors [1], the study of the groundwater-surface interaction systems [2], and heat or moisture transport in buildings [3]. These phenomena are typically modeled around an operating point by linear parabolic partial differential equations (PDEs), commonly known as diffusion-transport-reaction equations [4]. In most physical diffusion-transport-reaction systems the -exact- parameters are often unknown. Hence, data-driven estimation of the unknown physical parameters is necessary to determine a model which can be used for simulation, analysis, prediction, and control.

Historically, the utilization of measured data to determine an estimate of -physical- parameters is known as an inverse problem [5], [6]; in contrast to forward problems where

the model is used to generate data. Specifically, determining unknown physical parameters in dynamical systems is often considered as an optimization problem where an error criterion based on the mismatch between model and measurements is minimized. To this end, there are typically two sub-fields in PDE estimation: 1) grey-box identification where unknown parameters are estimated by minimizing the difference between a pre-selected model (class), e.g., a PDE with unknown diffusion, and measured data [7], [8]; 2) parameter learning which resorts to machine learning like techniques to avoid pre-selection of model-class [9], [10]. In both cases the underlying infinite dimensional models are generally approximated by finite dimensional models.

The standard method to estimate the unknown parameters is the output error criterion, i.e., taking the -weighted- sum of the squared error between the measurements and model output, primarily solved with iterative optimization methods [6]–[8], [11], [12]. These suffer from two problems: 1) they are solved iteratively, which is time consuming and in many cases the problem is non-convex, i.e., no guarantee for convergence to an optimal solution; 2) as only the output error is optimized and the unknown parameters are unconstrained, the parameters can start oscillating even when the output error is zero due to spatial aliasing of the state, which is estimated simultaneously. The latter is generally resolved by regularizing the unknown parameters [13]. However, this regularisation is often artificial, as there is usually no *a priori* information on how the unknown parameters change as function of space. Hence, here we propose a different approach by separating the two problems. First, we estimate the solution of the state over space based on a finite number of measurements. This allows us to transform the output error in an equation error criterion. By transforming this criterion into the frequency domain, we can derive a closed-form expression for the unknown parameters based on the estimated states, which is the novelty of this letter. Hence, we can uniquely and directly -without iteration- determine the diffusivity, convectivity, reactivity, and the source -as function of space-, simultaneously. Combining this with advances in 1) modern frequency domain signal processing to reduce noise and removing the initial condition [14], [15]; 2) recent innovations in dealing with experimentally unknown boundary conditions [16]; 3) a wide variety of functions to spatially parameterize transport [17]. This results in a highly versatile and fast method to acquire reliable estimates of the spatially varying parameters. This is in strong contrast to direct solutions proposed in the literature which are based on (piecewise) constant parameters [18] and often only consider diffusion [19], [20].

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II. PROBLEM FORMULATION

The estimation of space-dependent physical parameters in heat and mass transport is performed based on the following specifications.

a) *Model-Class*: A class of linear parabolic PDEs is considered in a one dimensional bounded spatial geometry to model the spatio-temporal dynamics of heat and mass transport. For all $x \in \mathbb{X} := [x_b, x_e] \subset \mathbb{R}$ and $t \in \mathbb{T} := [t_0, \infty) \subseteq \mathbb{R}_{\geq 0}$, the class of PDEs is defined as

$$\frac{\partial z}{\partial t} = D(x) \frac{\partial^2 z}{\partial x^2} + V(x) \frac{\partial z}{\partial x} + K(x)z + P(x)s(t). \quad (1)$$

Here, the state $z : \mathbb{X} \times \mathbb{T} \rightarrow \mathbb{R}$ is a multi-variable function (that can describe temperature or mass concentration) and (1) is understood point-wise in $x \in \mathbb{X}$ and $t \in \mathbb{T}$ with z evaluated as $z(x, t)$. The physical transport parameters are diffusivity $D : \mathbb{X} \rightarrow \mathbb{R}_{>0}$, convectivity $V : \mathbb{X} \rightarrow \mathbb{R}$, and reactivity $K : \mathbb{X} \rightarrow \mathbb{R}$. The external input is denoted by $s : \mathbb{T} \rightarrow \mathbb{R}$. Furthermore, the spatial distribution of the input is given by the function $P : \mathbb{X} \rightarrow \mathbb{R}$.

For well-posedness, the PDE is constrained by two boundary conditions at locations x_b and x_e . Moreover, the initial condition $z(\cdot, t_0)$ is assumed compatible with the model and its boundaries.

b) *Measured Data*: Corresponding to the heat and mass transport phenomena, location specific values of the state function z are measured over time and available as data. Let there be $M > 0$ sensors that measure z at the locations given by the set of points $\mathbb{X}_M := \{\check{x}_1, \check{x}_2, \dots, \check{x}_M\} \subset \mathbb{X}$. The measured output signals are $y^m(t) = z(\check{x}_m, t)$ for all $t \in \mathbb{T}$, with $m \in \{1, \dots, M\}$. The input $s(t)$ is assumed to be known or measured for all $t \in \mathbb{T}$.

c) *Problem Formulation*: In practical applications, the explicit definitions of space-dependent transport parameters $\{D(x), V(x), K(x), P(x)\}$ are often not available. Moreover, the boundary conditions that constrain heat and mass transport phenomena may depend on the spatially varying transport parameters, and, hence, remain unknown. In a similar fashion, in practice, the -exact- initial condition $z(\cdot, t_0)$ may also be unknown. Therefore, this letter takes the extremum measurements as boundary inputs, however, other linear boundary conditions are also allowed.

To complete the model in (1), the spatially varying transport parameters $\{D(x), V(x), K(x), P(x)\}$ have to be estimated based on measured data. This results in the following estimation problem.

Problem 1: Given the pre-processed data-set (see Remark 2)

$$\mathbb{D} := \{y^1(t), \dots, y^M(t), s(t) \mid t \in [t_0, \infty)\},$$

estimate the unknown function $\gamma : \mathbb{X}_E \rightarrow \mathbb{R}^4$,

$$\gamma(x) := \text{col}(D(x), V(x), K(x), P(x)), \quad (2)$$

by minimizing a cost function $\mathcal{V}(\mathbb{D}, z(x, t; \gamma))$ -defined in Section IV- over γ such that the solution $z(x, t; \gamma)$ satisfies the model (1) with parameters γ in the sense that

$$\frac{\partial z}{\partial t} = \text{row} \left(\frac{\partial^2 z}{\partial x^2}, \frac{\partial z}{\partial x}, z, s \right) \gamma, \quad (3)$$

subject to the boundary conditions and initial condition

$$z(\check{x}_1, t) = y^1(t), \quad z(\check{x}_M, t) = y^M(t), \quad z(x, t_0) = 0.$$

Remark 1: Using the extremum measurements as Dirichlet boundary conditions reduces the estimation domain to $\mathbb{X}_E := [\check{x}_1, \check{x}_M]$ [16]. Moreover, this allows to set-up the parameter estimation problem without the need of full knowledge about the actual boundary conditions.

Remark 2: With -advanced- signal processing techniques, the measured signal can be split into a transients/drift signal (non-steady-state behavior, e.g., from the initial condition), a forced response (from the excitation) and additive (filtered) noise [14], [15]. By removing the transient and noise terms from the original signal, only the forced response remains in the filtered data set \mathbb{D} , which is equivalent to $z(\cdot, t_0) = 0$.

III. FINITE DIMENSIONAL FREQUENCY DOMAIN PROBLEM

Problem 1 is infinite dimensional with no known analytic solution for $z(x, t)$. Therefore, this section approximates the infinite dimensional problem by creating a finite dimensional model following the methodology from [17].

A. Frequency Domain Approach

Assuming that the discrete Fourier transformed input $S(k)$ has (excited) frequency bins $k \in \mathbb{K}$, Problem 1 can be studied in the frequency domain without loss of information (Parseval's Theorem). Moreover, due to linearity of the model, $Z(x, k)$ is independent for each excited bin k . The frequency domain model of (3) is then given by

$$\begin{aligned} i\omega_k Z &= \text{row} \left(\frac{\partial^2 Z}{\partial x^2}, \frac{\partial Z}{\partial x}, Z \right) \gamma \\ Y(k) &:= \text{col}(Z(\check{x}_2, k), \dots, Z(\check{x}_{M-1}, k)) \end{aligned} \quad (4)$$

subject to the boundary conditions

$$Z(\check{x}_1, k) = Y^1(k), \quad Z(\check{x}_M, k) = Y^M(k),$$

with discrete Fourier transformed state $Z : \mathbb{X} \times \mathbb{K} \rightarrow \mathbb{C}$, input $S : \mathbb{K} \rightarrow \mathbb{C}$, output $Y(k)$, $i^2 = -1$ and angular frequency ω_k corresponding to the k^{th} -bin.

Remark 3: In excitation experiments only a finite number of bins are informative, i.e., those bins which are present in the input $S(k)$ and are above the noise level [14]. Hence, in practice only a -few- finite number of bins need to be considered (see [21] for details).

B. Linear Parameterization of the Unknown Functions

For estimation purposes, assume that γ belongs to a function space Γ that is parameterized by a surjective mapping $\Pi : \Theta \rightarrow \Gamma$ that is described by a finite sum of basis functions $B_r(x) := \text{diag}(B_r^D(x), B_r^V(x), B_r^K(x), B_r^P(x))$,

$$\gamma(x; \theta) := [\Pi(\theta)](x) := \sum_{r=1}^R B_r(x) \theta_r, \quad (5)$$

with $\theta = \text{col}(\theta_1, \dots, \theta_R)$ and $\theta_r = \text{col}(\theta_r^D, \theta_r^V, \theta_r^K, \theta_r^P) \in \Theta \subset \mathbb{R}^4$. With this parameterization, the estimation of γ amounts to estimating $\theta \in \mathbb{R}^{4R}$.

C. Semi-Discretization

The infinite dimensional model (4) is approximated by a central finite difference scheme that converges to the exact solution for $N \rightarrow \infty$ [22]. The finite dimensional model is

$$\begin{aligned} i\omega_k \mathbf{Z}(k) &= A(\theta)\mathbf{Z}(k) + B(\theta)U(k) \\ Y(k) &= C\mathbf{Z}(k) \end{aligned} \quad (6)$$

with state vector $\mathbf{Z}(k) := \text{col}(Z(x_2, k), \dots, Z(x_{N-1}, k))$ at sample $x_j \in \mathbb{X}_d \subset \mathbb{X}$, $j \in \{1, \dots, N\}$, extended input vector $U(k) = \text{col}(S(k), Y^1(k), Y^M(k))$. Here, $A(\theta)$ and $B(\theta)$ contain the boundary conditions, are linear affine in θ , and defined in [17]. The -observation- matrix C maps states to output. Altogether, the finite dimensional problem yields.

Problem 2: Given the pre-processed data-set

$$\mathbb{D} = \{\text{col}(Y^1(k), \dots, Y^M(k)), S(k) \mid k \in \mathbb{K}\}.$$

Estimate θ by minimizing a cost function $\mathcal{V}(\mathbb{D}, \mathbf{Z}(k; \theta))$ over θ such that $\mathbf{Z}(k; \theta)$ satisfies the model (6).

IV. THE INVERSE PROBLEM

This section describes our methodology to estimate the unknown weights. In the literature for distributed parameter systems, different error criteria are proposed to calculate the parameters. The commonly used criterion is the output error criterion [6]–[8], [11], [17] and the rarely used equation error criterion [7] which we will use in the frequency domain to derive a closed-form solution for the unknown parameters.

A. Output Error Criterion

The commonly used cost function is the output error criterion which is the sum of the squared error between solution $z(x, t; \gamma)$ for a given γ and measurements $y(t)$, i.e.,

$$\mathcal{V}_{\text{oe}}(\theta) := \int_{t_0}^{\tau} \int_{x_b}^{x_e} |y(t) - \mathcal{C}z(x, t; \gamma(x; \theta))|^2 dx dt \quad (7)$$

with observation map \mathcal{C} that maps state to output. This criterion is transformed into the frequency domain such that the output error criterion for Problem 2 and [17] is given by

$$\mathcal{V}_{\text{oe}}(\theta) = \sum_{k \in \mathbb{K}} \left\| Y_k - C(i\omega_k I - A(\theta))^{-1} B(\theta) U_k \right\|^2, \quad (8)$$

with shorthand notation $Y_k = Y(k)$ and $U_k = U(k)$. This criterion is nonlinear in θ due to the inverse and multiplication. As such, it needs to be iteratively optimized with often no guarantee for convergence to the global minimum. As this is generally the case for the output error criterion, we propose to use the equation error criterion for which a closed-form solution can be derived.

B. Equation Error Criterion

In the equation error criterion, the state of the model is replaced by measurements or estimates of the state $\hat{z}(x, t)$ [7], such that for (3), the equation error criterion is defined as

$$\begin{aligned} \mathcal{V}_{\text{ee}}(\theta) &:= \int_{t_0}^{\tau} \int_{x_1}^{x_M} \left(\frac{\partial \hat{z}}{\partial t} - \text{row} \left(\frac{\partial^2 \hat{z}}{\partial x^2}, \frac{\partial \hat{z}}{\partial x}, \hat{z}, s \right) \gamma(x; \theta) \right)^2 dx dt. \quad (9) \end{aligned}$$

This can be simplified in the frequency domain: the integral over time simplifies to a summation over the excited frequency

bins $k \in \mathbb{K}$ without loss of information (Parseval's theorem). With the measured or estimated state vector in the frequency domain $\hat{\mathbf{Z}}_k$, the equation error for Problem 2 simplifies to

$$\mathcal{V}_{\text{ee}}(\theta) = \sum_{k \in \mathbb{K}} \left\| i\omega_k \hat{\mathbf{Z}}_k - \left(A(\theta)\hat{\mathbf{Z}}_k + B(\theta)U_k \right) \right\|^2. \quad (10)$$

C. Derivation of the Closed-Form Solution

Consider the equation error criterion in (10) and that $A(\theta)$ and $B(\theta)$ are linear affine in θ , then new matrices \bar{A} , \bar{B} can be defined as a function of the data U_k and $\hat{\mathbf{Z}}_k$ such that $A(\theta)\hat{\mathbf{Z}}_k = \hat{A}(\hat{\mathbf{Z}}_k)\theta$ and $B(\theta)U_k = \hat{B}(U_k)\theta$ (see the Appendix). As a result, (10) can be written as

$$\mathcal{V}_{\text{ee}} = \sum_{k \in \mathbb{K}} \left\| i\omega_k \hat{\mathbf{Z}}_k - \left(\hat{A}(\hat{\mathbf{Z}}_k) + \hat{B}(U_k) \right) \theta \right\|^2, \quad (11)$$

with the closed-form solution

$$\hat{\theta} := \left(\bar{A} + \bar{B} \right)^{\text{H}} \left(\bar{A} + \bar{B} \right)^{-1} \left(\bar{A} + \bar{B} \right)^{\text{H}} \bar{W}, \quad (12)$$

where H denotes the Hermitian transpose, \bar{A} , \bar{B} , and \bar{W} are the column concatenation of $\hat{A}(\hat{\mathbf{Z}}_k)$, $\hat{B}(U_k)$, and $i\omega_k \hat{\mathbf{Z}}_k$ for all excited bins $k \in \mathbb{K}$, respectively. Hence, the optimal weighting for θ , and thus $\gamma(x)$, are determined without using iterative optimization methods.

Remark 4: The equation error in combination with different cost functions such as weighted and total linear least squares also have closed-form solutions [23].

D. State Estimation

The unique solution (12) requires knowledge of the state at each discretization point, i.e., a “space-”continuous measurement. If $N \rightarrow \infty$, the finite dimensional description converges to the true infinite dimensional solution and an exact solution can be found.

However, due to limited spatial measurements, the purpose is to estimate the full state based on these measurements. This raises the fundamental problem of unknown in-between sensor behavior. In case of regularized output error [7], the intermediate relationship in-between measurements is determined by the regularized functions for the transport parameters. This technique is reminiscent to the (spatial) Nyquist Shannon (NS) sampling theorem. Therefore, a consistent -spatial- signal reconstruction using the NS theorem is required and currently being worked out. Alternatively, the states can be inferred by interpolating the measurements using machine learning techniques -currently being further worked out- [24], [25]; classic interpolation methods; or reducing the number of states in the model to match the measurements. Note that the selected method should be model free such that the problem remains affine in the unknown parameters. In the simulation section, we show the latter two approximation methods in combination with the closed-form solution.

V. SIMULATION RESULTS

In this section, two simulation scenarios are presented demonstrating the merit of the proposed methodology. As the here proposed method has a closed-form solution that only requires solving a linear matrix equality, it is expected to be significantly faster than iterative optimization methods, e.g.,

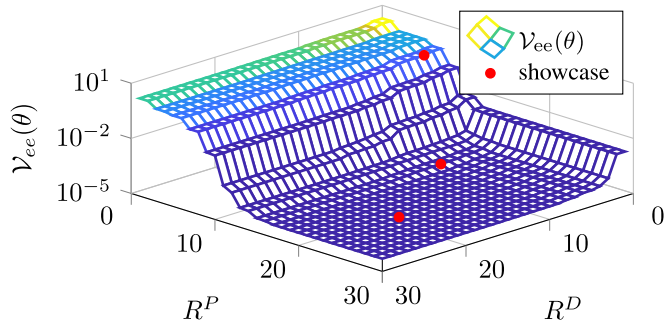


Fig. 1. The fitting error map $V_{ee}(\theta)$ for scenario 1 by approximating the parameters $\{D(x), P(x)\}$ with different number of basis-functions, R^D , R^P , respectively. Here, the \bullet annotates the locations $(R^D, R^P) = (3, 8)$, $(9, 16)$, and $(21, 23)$ shown in Fig. 2.

when compared to [17]. This can be exploited by considering two scenarios: *scenario 1* which estimates $\{D(x), P(x)\}$ testing different orders (weights) R of the basis functions and *scenario 2* which tests different interpolation density of the data when all four functions $\{D(x), V(x), K(x), P(x)\}$ are estimated using fixed set of basis functions $B_r(x)$ and their orders R .

A. Data Generation

The simulation example is inspired by perturbative experiments in the field of nuclear fusion [26]. The heat transport, (1), is generally analyzed on the normalized domain $\mathbb{X} = [x_b, x_e] = [0, 1]$ of the -minor- plasma radius. Here x_b is at the center and x_e is at the edge of the plasma. The corresponding boundary conditions in the simulation are $\frac{\partial Z}{\partial x}(x_b, \cdot) = 0$ due to (axi)symmetry and Dirichlet boundary condition $Z(x_e, \cdot) = 0$ due to a significant temperature difference between core plasma 170 million °C and edge plasma 1 million °C. Typical functions used in nuclear fusion are $D^{\text{sim}}(x) = 5x^3 - 0.005x + 5$, $V^{\text{sim}}(x) = 15x^2 - 0.005$, $K^{\text{sim}}(x) = -3x$, and $P^{\text{sim}}(x) = 0.2 + \frac{7}{\sqrt{\pi}} \exp\left(\frac{-(x-0.35)^2}{(0.1)^2}\right) + \frac{5.6}{\sqrt{\pi}} \exp\left(\frac{-(x-0.6)^2}{(0.1)^2}\right)$. For the perturbation of the plasma temperature a microwave source is used, where the excitation signal $S(\omega)$ is a block-wave of $\omega_0 = 50\pi$ with a 70% duty cycle. Here, only the first five harmonics $\omega_k = k\omega_0$, $k = 1, \dots, 5$ have a significant contribution and are used for the estimation. The temperature data is generated by a simulation with a central finite difference grid of $N = 801$ sample points.

B. Estimation of $\{D(x), P(x)\}$ With Unknown Order R

In *scenario 1*, the goal is to estimate $\{D(x), P(x)\}$ without knowledge on the correct orders (R^D, R^P) in (5), which practically means we do not know the shape of $\{D(x), P(x)\}$. For ease of explanation, we set $V(x) = K(x) = 0$ for this example. As the number of sensors plays an important role, we choose here $M = 22$ sensors that are located at $\tilde{x}_m = 0.05 + 0.04125(m-1)$, with $m = 1, \dots, M$. This corresponds to the electron cyclotron emission (ECE) diagnostic that measures in a medium sized fusion reactor.

In principle, any basis function can be used and several can be tested simultaneously. Here, we use -arbitrarily chosen-Chebyshev polynomials for both parameters [22].

We vary the orders $R^D, R^P \in \{1, \dots, 30\}$ and use a discretization grid that equals the measurement grid $\mathbb{X}_d = \mathbb{X}_M$,

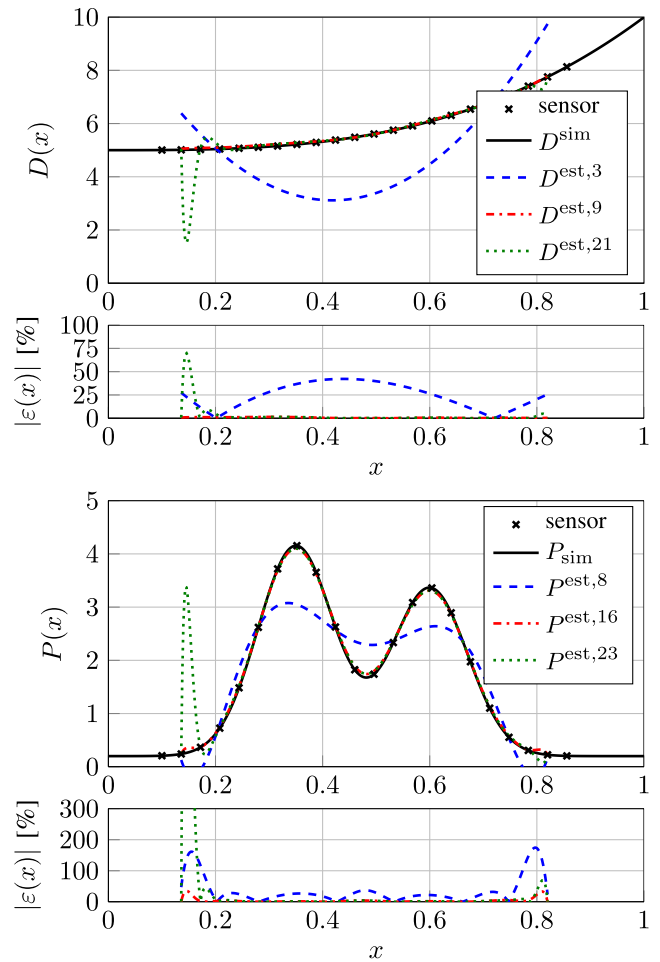


Fig. 2. The estimated parameters $\{D(x), P(x)\}$ over \mathbb{X}_E for scenario 1 with the number of basis-function $(R^D, R^P) = (3, 8)$, $(9, 16)$ and $(21, 23)$ and their relative error $\varepsilon(x)$.

thus $C = I$. Naturally, the cost (10) decreases for increasing R^D and R^P as shown in Fig. 1. For three different combinations, i.e., $(R^D, R^P) = (3, 8)$, $(9, 16)$, and $(21, 23)$ the resulting estimates are shown in Fig. 2.

Fig. 2 shows that when the order (R^D, R^P) is too low, e.g., for $(3, 8)$, significant errors occur in the estimates. If the order is sufficiently high, e.g., for the pairs $(9, 16)$ and $(21, 23)$, the estimates have the correct value at the sensor locations as is imposed by the equation error. Consequently, only the values of $\gamma(x)$ at the sensor locations should be considered as a correct result given that the combination of basis functions and order give sufficient freedom.

Remark 5: Although the intermediate points between sensor locations should not be considered in this methodology, $\gamma(x)$ is defined at these points and plotted for completeness. As the comparison of the orders $(9, 16)$ and $(21, 23)$ shows, the finite dimensional approximation does not pose a unique solution for $\gamma(x)$ at the intermediate points. This can be resolved by increasing the discretization grid, e.g., by interpolating the data, which is further investigated in the following section.

C. Estimation Using Spatial Interpolations of the Temperature

In the simulation for *scenario 2*, we investigate the effect of interpolations on the estimation. In *scenario 2* all the functions

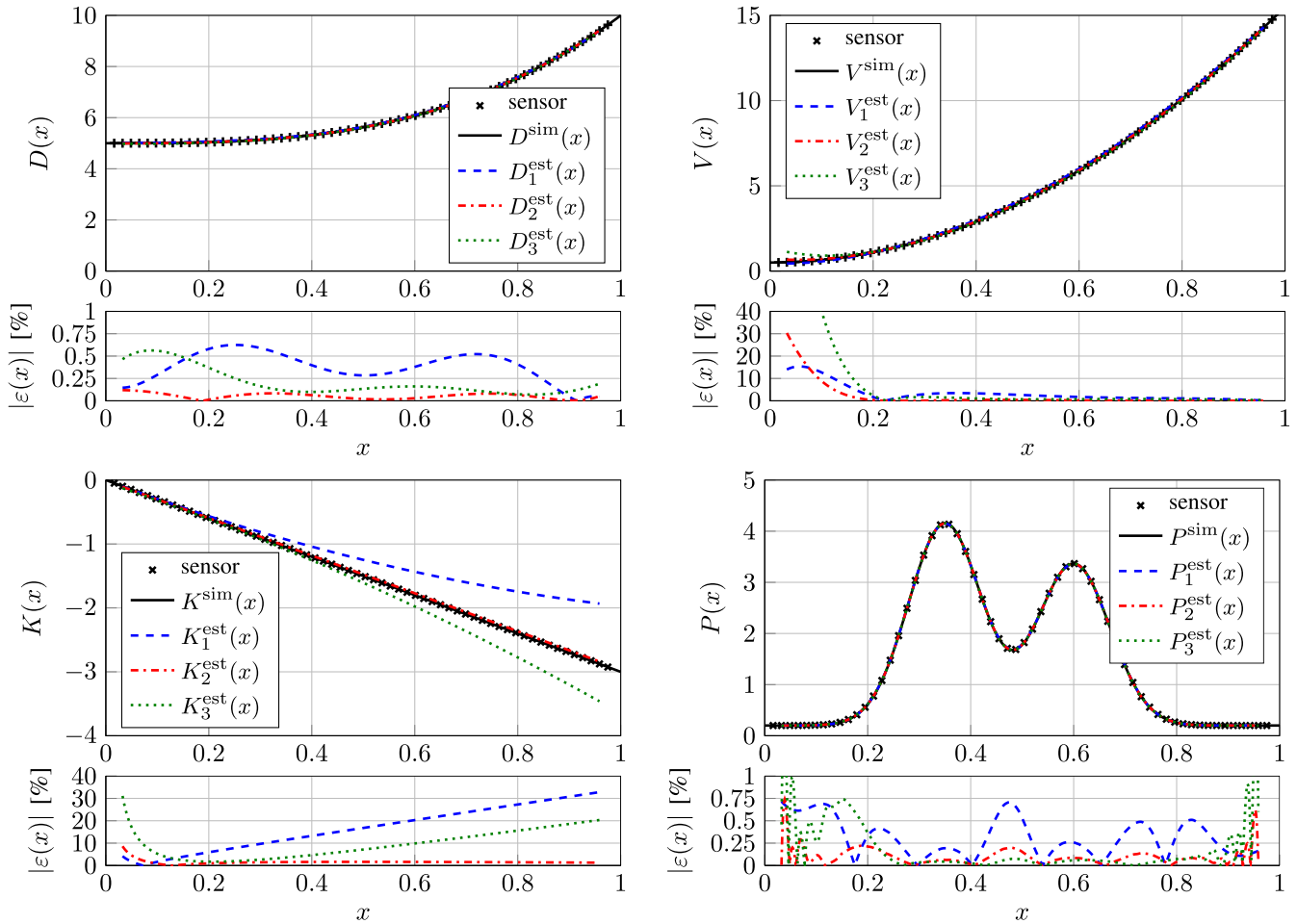


Fig. 3. Estimated parameters $\{D(x), V(x), K(x), P(x)\}$ over \mathbb{X}_E for scenario 2 and the relative error $\varepsilon(x)$ between the simulated parameter and the estimated parameter. The subscript 1, 2, 3 denote if the estimates are based on the measurements $N_1 = M$, or the interpolated measurements $N_2 = 2(M - 1) + 1$ or $N_3 = 10(M - 1) + 1$, respectively.

are non-zero and according to Section V-A. Here, we use a sensor grid which corresponds to that of a larger fusion reactor, i.e., the temperature is measured by $M = 60$ sensors located at $\tilde{x}_m = 0.1625m$, with $m = 1, \dots, M$.

In this scenario, the order of the basis functions is fixed and only the weights are estimated. Polynomial basis functions are used for $\{D(x), V(x), K(x)\}$ and B-spline functions for $P(x)$ to approximate Gaussians. The choice of order for the monomial basis functions to estimate $\{D(x), U(x), K(x)\}$ is higher than the actual order and is chosen to be $R^D = 8$, $R^V = 6$ and $R^K = 3$ such that it is possible to find an exact description. The basis function for $P(x)$, $B_r^B(x)$ is designed using the De Boor's algorithm [27] with 58 equally distributed control points, thus $R^P = 58$.

Three estimations are presented where, 1) the discretization grid equals the measurement grid $N_1 = M$, 2) the measurement data is interpolated over space to generate artificial spatial measurements, i.e., $N_2 = 2(M - 1) + 1$, and 3) more interpolation points, i.e., $N_3 = 10(M - 1) + 1$. The data is interpolated using cubic splines.

The estimation results are shown in Fig. 3. Overall, the estimated parameters $\{D(x), P(x)\}$ closely match with the simulated parameters, while $\{V(x), K(x)\}$ are estimated with a significantly lower accuracy. Estimating $\{V(x), K(x)\}$ requires a higher accuracy of the discretization grid. Therefore, interpolating the data increases the accuracy (see results for N_2),

however, over-interpolation can affect the estimation accuracy negatively (see results for N_3).

VI. CONCLUSION AND DISCUSSION

This letter presents a novel method to efficiently estimate the unknown space-dependent transport parameters, based on a closed-form solution of the equation error criterion. The closed-form solution is formulated as a linear matrix equality such that high density grids can be solved computationally efficient. As a result, if the states $z(\cdot, t)$ are known, the parameters are estimated uniquely with the desired accuracy by taking a sufficiently dense discretization grid.

In practice, measurements are often only provided at a limited set of spatial locations. This is generally resolved by (i) applying regularization or restrictions on the unknown parameters or (ii) as we have done here by “interpolating” the measurements to increase the grid density. There are three reasons why we prefer approach (ii): 1) As we are estimating the unknown parameters, prior information on the -smoothness of- parameters is generally unavailable, whereas smoothness on the states is required due to the underlying parabolic PDE; 2) the states as function of space can easily be -visually- inspected for correctness and validated by taking additional spatial measurements; and 3) a closed-form solution can be used which significantly speeds up the process and avoids convergence to a local minimum.

APPENDIX

The matrices used in (11) are given by

$$\begin{aligned}\hat{\mathbf{A}}(\hat{\mathbf{Z}}(k)) &:= [L_1^D, L_1^V, L_1^K, \mathbf{0}, \dots, \\ &\quad L_R^D, L_R^V, L_R^K, \mathbf{0}](I_{4R} \otimes \hat{\mathbf{Z}}(k)), \\ \hat{\mathbf{B}}(U(k)) &:= [\mathbf{0}, g_1^D, h_1^D, \mathbf{0}, g_1^V, h_1^V, \mathbf{0}, \mathbf{0}, \mathbf{0}, f_1^P, \mathbf{0}, \mathbf{0}, \dots, \\ &\quad \mathbf{0}, g_R^D, h_R^D, \mathbf{0}, g_R^V, h_R^V, \mathbf{0}, \mathbf{0}, \mathbf{0}, f_R^P, \mathbf{0}, \mathbf{0}] \\ &\quad (I_{4R} \otimes U(k)),\end{aligned}$$

where \otimes denotes the Kronecker product and $\mathbf{0}$ the zero vector/matrix of the appropriate size. The central finite difference matrices L with grid sample $\Delta_x > 0$ are

$$\begin{aligned}L_r^D &:= \frac{1}{(\Delta_x)^2} \tilde{B}_r^D \begin{bmatrix} -2 & 1 & & & \\ 1 & -2 & 1 & & \\ & \ddots & \ddots & \ddots & \\ & & & 1 & -2 \\ & & & & \end{bmatrix}, \\ L_r^V &:= \frac{1}{2\Delta_x} \tilde{B}_r^V \begin{bmatrix} 0 & 1 & & & \\ -1 & 0 & 1 & & \\ & \ddots & \ddots & \ddots & \\ & & & \ddots & \\ & & & & -1 & 0 \end{bmatrix}, \quad L_r^K := \tilde{B}_r^K.\end{aligned}$$

Here, \tilde{B}_r^D , \tilde{B}_r^V and \tilde{B}_r^K are diagonal matrices of dimension $(N-2) \times (N-2)$ with the diagonal entries $B_r^D(x_j)$, $B_r^V(x_j)$, $B_r^K(x_j)$ evaluated at each grid point x_j , $j \in \{2, \dots, N-1\}$. The vectors for the input and boundary conditions are $g_r^D := \text{col}(\frac{B_r^D(x_2)}{(\Delta_x)^2}, 0, \dots, 0)$, $g_r^U := \text{col}(-\frac{B_r^V(x_2)}{2\Delta_x}, 0, \dots, 0)$, $h_r^D := \text{col}(0, \dots, 0, \frac{B_r^D(x_{N-1})}{(\Delta_x)^2})$, $h_r^U := \text{col}(0, \dots, 0, \frac{B_r^V(x_{N-1})}{2\Delta_x})$, and $f_r^P := \text{col}(B_r^P(x_2), \dots, B_r^P(x_{N-1}))$.

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