

## Converses for Write-Unidirectional Memories

***Citation for published version (APA):***

Willems, F. M. J. (1989). *Converses for Write-Unidirectional Memories*. (EUT report. E, Fac. of Electrical Engineering; Vol. 89-E-220). Eindhoven University of Technology.

***Document status and date:***

Published: 01/01/1989

***Document Version:***

Publisher's PDF, also known as Version of Record (includes final page, issue and volume numbers)

***Please check the document version of this publication:***

- A submitted manuscript is the version of the article upon submission and before peer-review. There can be important differences between the submitted version and the official published version of record. People interested in the research are advised to contact the author for the final version of the publication, or visit the DOI to the publisher's website.
- The final author version and the galley proof are versions of the publication after peer review.
- The final published version features the final layout of the paper including the volume, issue and page numbers.

[Link to publication](#)

***General rights***

Copyright and moral rights for the publications made accessible in the public portal are retained by the authors and/or other copyright owners and it is a condition of accessing publications that users recognise and abide by the legal requirements associated with these rights.

- Users may download and print one copy of any publication from the public portal for the purpose of private study or research.
- You may not further distribute the material or use it for any profit-making activity or commercial gain
- You may freely distribute the URL identifying the publication in the public portal.

If the publication is distributed under the terms of Article 25fa of the Dutch Copyright Act, indicated by the "Taverne" license above, please follow below link for the End User Agreement:

[www.tue.nl/taverne](http://www.tue.nl/taverne)

***Take down policy***

If you believe that this document breaches copyright please contact us at:

[openaccess@tue.nl](mailto:openaccess@tue.nl)

providing details and we will investigate your claim.



Research Report

ISSN 0167-9708

Coden: TEUEDE

Eindhoven  
University of Technology  
Netherlands

Faculty of Electrical Engineering

# Converses for Write- Unidirectional Memories

by  
Frans M.J. Willems

EUT Report 89-E-220  
ISBN 90-6144-220-6

May 1989

**Eindhoven University of Technology Research Reports  
EINDHOVEN UNIVERSITY OF TECHNOLOGY**

**Faculty of Electrical Engineering  
Eindhoven The Netherlands**

ISSN 0167-9708

Coden : TEUEDE

## **CONVERSES FOR WRITE-UNIDIRECTIONAL MEMORIES**

by

**Frans M.J. Willems**

**EUT Report 89-E-220  
ISBN 90-6144-220-6**

**Eindhoven  
May 1989**

CIP-GEGEVENS KONINKLIJKE BIBLIOTHEEK, DEN HAAG

Willems, F.M.J.

Converses for write-unidirectional memories / by F.M.J. Willems. - Eindhoven :  
Eindhoven University of Technology, Faculty of Electrical Engineering. - Tab. -  
(EUT report, ISSN 0167-9708; 89-E-220)

Met lit. opg., reg.

ISBN 90-6144-220-6

SISO 520.6 UDC 681.327.68:519.72 NUGI 832

Trefw.: optische geheugens; informatietheorie.

## Contents

	Abstract . . . . .	1
1.	Introduction . . . . .	1
2.	Definitions . . . . .	2
3.	Results . . . . .	5
4.	Proof of Lemma 1 . . . . .	6
5.	Proof of Lemma 2 . . . . .	8
6.	Conclusions . . . . .	11
7.	Acknowledgement . . . . .	11
	References . . . . .	11

---

# CONVERSES FOR WRITE-UNIDIRECTIONAL MEMORIES

Frans M.J. Willems\*

**ABSTRACT :** First we show that we can transform a WUM channel into a channel that behaves identical in even and odd cycles. For this derived channel we then prove that the average-error capacity cannot exceed  $\log_2((1+\sqrt{5})/2) = 0.69424\dots$ . For the average-error capacity of the derived channel in the situation where both encoder and the decoder are uninformed, we find an even better upper bound (0.54588\dots). In this case however we assume that in all cycles the same code is used.

## 1. Introduction

In 1985, researchers at Philips Research Laboratories were interested in the capacity aspects of a magneto-optical recording system [1]. In this system a laser beam is used to heat up a spot on an optical disc. Depending on the orientation of the magnetic field, either a zero or a one will be written on this spot. The magnetic field is generated by an electromagnet. An optical effect makes it possible to retrieve the information stored in a spot.

A problem arises when we want to record information at high speed. The inductivity of the electromagnet will prevent us from reversing the current too often. Therefore the following strategy is proposed. Suppose a new disc contains only zeros. During the first cycle we can store information on the disc by changing some of the zeros into ones by switching the laser on and off, without having to change the polarity of the current through the electromagnet. In the second cycle we reverse the polarity and we restrict ourselves to writing only zeros, again by switching the laser on and off, and keeping the remaining components unchanged. In the third cycle we write only ones, etc.

An additional feature of the described recording system is that before each cycle the writer knows the state of the disc. This side information may be used to increase the efficiency of the coding process. The reader however is assumed not to be aware of the previous state of the disc.

Philips researchers developed simple (time-sharing) codes that achieved a rate of 0.5 bit per spot and were interested in codes with higher rates. In the "Applications"-

---

\* Eindhoven University of Technology, Faculty of Electrical Engineering, Information Theory Group, P.O. Box 513, 5600 MB Eindhoven, The Netherlands.

session at the Seventh Symposium on Information Theory in the Benelux in 1986 Willems and Vinck [2] presented the first article in this field. They found a simple code with a rate of 0.51699 bit/spot, thereby beating the 0.5 barrier. Willems and Vinck noted that there was a connection between the magneto-optical recording channel and the Blackwell broadcast channel (see e.g. Gel'fand [3] and Pinsker [4]), and recognizing this, they stated (however without proof) that  $\log_2((1+\sqrt{5})/2) = 0.69424$  is the average-error capacity of the magneto-optical recording channel.

In 1986 Borden [5] investigated a slightly different system which he named "Write-Unidirectional Memory (WUM)". Instead of zeros in even cycles and ones in odd cycles, in Borden's model the encoder is allowed in each cycle, after having inspected the (previous) state of the disc, to choose to write either zeros or ones. Borden showed that 0.69424 bit/spot is the capacity of the WUM-channel in the zero-error case.

Subsequently Simonyi [6] generalized the WUM-model by observing that both the writer and the reader could be either "informed" or "uninformed" of the previous state of the channel (disc). In this way he obtained four different models. He assumed that the current through the electromagnet is reversed at the beginning of each new cycle as in Willems and Vinck [1]. For the "classical" WUM-channel (encoder informed, decoder uninformed) Simonyi generalized the code of Willems and Vinck and found a code with a rate of 0.53254 bit/spot.

In this report we will give the weak converses for the four models that are described by Simonyi. We assume that the error probability concept is the average-error concept. The current through the electromagnet is assumed to be reversed each cycle.

## 2. Definitions

Let  $N \in 1, 2, \dots$ . A WUM of block length  $N$  consists of  $N$  components  $y^N := (y_1, y_2, \dots, y_N)$ . Each component  $y_n$ ,  $n \in \{1, 2, \dots, N\}$  may assume a value from  $\{0, 1\}$ .

A decoder can inspect the components of the memory. An encoder is a device that can alter them.

A cycle (indexed by  $k \in 1, 2, \dots$ ) is a time interval that starts when the  $k$ -th "message" is stored in the WUM and that ends when message  $k+1$  is about to be stored. During odd cycles the encoder can decide to leave a component unchanged, we say the encoder writes a "?" or to write a "1". In an even cycle a component can remain unaltered, a "?" is written, or can be set equal to "0". The tables below give

the updated value  $y_n(k)$  of a WUM-component given its previous value  $y_n(k-1)$  and the "input"  $f_n(k)$ .

$f(k)$	$y(k-1)$	$y(k)$
?	0	0
?	1	1
1	0	1
1	1	1

k odd

$f(k)$	$y(k-1)$	$y(k)$
0	0	0
0	1	0
?	0	0
?	1	1

k even

Table 1.  
 $y(k)$  as a function of  $y(k-1)$  and the input  $f(k)$  for both odd and even  $k$  ( $n$  fixed).

At the beginning of cycle  $k$  we denote the state of the WUM by  $y^N(k) = (y_1(k), y_2(k), \dots, y_N(k))$ . An information source generates the message  $w(k) \in \{0, 1, \dots, M-1\}$ . We assume that  $\Pr\{W(k)=m\} = 1/M$  for  $m \in \{0, 1, \dots, M-1\}$ . The encoder maps the message  $w(k)$  into a "prescription"  $f^N(k) := (f_1(k), f_2(k), \dots, f_N(k))$ . When  $k$  is odd  $f_n(k) \in \{?, 1\}$ , when  $k$  is even  $f_n(k) \in \{0, ?\}$ . Yet we will see that the distinction between odd and even cycles is artificial. Therefore we introduce the random variables  $G_n(k)$  and  $Z_n(k)$  for  $k = (0, 1, 2, \dots)$ . When we define

$$G_n(k) := \begin{cases} i & \text{if } F_n(k) = ?, \\ r & \text{else,} \end{cases} \quad \text{for } k = 1, 2, \dots \text{ and} \quad (1a)$$

$$Z_n(k) := \begin{cases} 1 - Y_n(k) & \text{for } k \text{ odd,} \\ Y_n(k) & \text{for } k \text{ even,} \end{cases} \quad \text{for } k = 0, 1, 2, \dots, \quad (1b)$$

we see that the mapping that determines a component  $z_n(k)$  from the  $z_n(k-1)$  and  $g_n(k)$  does not depend on  $k$  anymore (see table 2).

$g(k)$	$z(k-1)$	$z(k)$
i	0	1
i	1	0
r	0	0
r	1	0

Table 2.  
 $z(k)$  as a function of  $z(k-1)$  and the input  $g(k)$  ( $n$  fixed).

As can be seen in the table, the combination  $(z_n(k), z_n(k-1)) = (1, 1)$  never

occurs. From now on we consider a WUM as a channel, where for each component there is an input  $g(k) \in \{i,r\}$ , state  $z(k-1) \in \{0,1\}$  and an output  $z(k) \in \{0,1\}$  that depends on  $g(k)$  and  $z(k-1)$  as in table 2.

We say that an encoder is "uninformed" when the previous state of the channel is not used in the determination of  $g^N(k)$ , hence

$$g^N(k) := E(w(k),k). \quad (2a)$$

The encoder is "informed" when

$$g^N(k) := E(w(k),z^N(k-1),k). \quad (2b)$$

Likewise a decoder is said to be "uninformed" when

$$\hat{w}(k) := D(z^N(k),k), \quad (3a)$$

and "informed" when

$$\hat{w}(k) := D(z^N(k),z^N(k-1),k). \quad (3b)$$

Note that both the encoder and decoder are allowed to use codes that depend on the cycle index  $k$ . For practical reasons however, we assume that there exists a "period"  $T \in \{1,2,\dots\}$  such that  $E(\cdot,k) = E(\cdot,k \bmod T)$  or  $E(\cdot,\cdot,k) = E(\cdot,\cdot,k \bmod T)$  and  $D(\cdot,k) = D(\cdot,k \bmod T)$  or  $D(\cdot,\cdot,k) = D(\cdot,\cdot,k \bmod T)$ . The entire coding scheme is now referred to as a "period- $T$ " code. E.g. if we use one odd-cycle code and one even-cycle code we have a period-2 code.

By combining (2a) or (2b) with (3a) or (3b) we find four different cases. For each of these cases we can design codes. The (average-) error probability of a code is

defined as

$$P_e := \lim_{K \rightarrow \infty} \frac{1}{K} \cdot \sum_{k=1, K} P_e(k), \text{ with } P_e(k) := \Pr\{\hat{W}(k) \neq W(k)\}, \quad (4)$$

where we assume that  $z^N(0) = 0^N$ .

A rate  $R$  is said to be  $T$ -achievable if for every  $\delta > 0$  there exist for all  $N$  large enough period- $T$  codes with  $\frac{1}{N} \cdot \log(M) \geq R - \delta$  and error probability  $P_e \leq \delta$ . The capacity  $C(T)$  is defined as the maximum of all  $T$ -achievable rates.

It will be clear now that there are four different WUM-capacities, possibly depending on the period  $T$ , if we consider the average-error concept.

The binary entropy function  $h(\cdot)$  is defined as

$$h(\alpha) := -\alpha \cdot \log(\alpha) - (1-\alpha) \cdot \log(1-\alpha), \text{ for } 0 \leq \alpha \leq 1. \quad (5)$$

Throughout this report we assume that the basis of the logarithm is 2.

### 3. Results

In this report we prove two converses :

**Lemma 1 :** For the WUM with informed encoder and informed decoder for arbitrary period  $T$

$$C_{\text{informed, informed}}^{(T)} \leq \log((1+\sqrt{5})/2). \quad \square$$

**Lemma 2 :** For the WUM with uninformed encoder and uninformed decoder for period  $T = 1$

$$C_{\text{uninformed, uninformed}}^{(T=1)} \leq \max_{\beta} (1-\beta) \cdot h(\beta/(1+\beta)). \quad \square$$

From Lemma 1 we immediately obtain

**Corollary 1** : For the WUM with only the encoder or only the decoder informed and arbitrary period T

$$C_{\text{uninformed, informed}}^{(T)} \leq \log((1+\sqrt{5})/2), \text{ and} \quad \square$$

$$C_{\text{informed, uninformed}}^{(T)} \leq \log((1+\sqrt{5})/2). \quad \square$$

#### 4. Proof of Lemma 1

For  $k = 1, 2, \dots$  the Fano-inequality yields

$$H(W(k) | \hat{W}(k)) \leq \phi(P_e(k)), \text{ where} \quad (6a)$$

$$\phi(p) \quad := h(p) + p \cdot \log(M-1), \text{ for } 0 \leq p \leq 1. \quad (6b)$$

Note that  $\phi(p)$  is convex in  $p$  and that  $\phi(p)/N \downarrow 0$  when  $p \downarrow 0$  ( $M$  constant).

Now for  $k = 1, 2, \dots$

$$\begin{aligned} \log(M) &= H(W(k)) = H(W(k) | Z^N(k-1)) \\ &= I(W(k); Z^N(k) | Z^N(k-1)) + H(W(k) | Z^N(k), Z^N(k-1)) \\ &\stackrel{(a)}{=} I(W(k); Z^N(k) | Z^N(k-1)) + H(W(k) | Z^N(k), Z^N(k-1), \hat{W}(k)) \\ &\leq H(Z^N(k) | Z^N(k-1)) + H(W(k) | \hat{W}(k)) \\ &\stackrel{(b)}{\leq} H(Z^N(k) | Z^N(k-1)) + \phi(P_e(k)). \end{aligned} \quad (7)$$

Here (a) follows from the fact that  $\hat{w}(k)$  is determined by  $z^N(k)$  and  $z^N(k-1)$  as stated in (3b), and (b) is Fano's inequality. Next we find

$$\begin{aligned} \frac{1}{N} \cdot \log(M) &\leq \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \sum_{n=1, N} H(Z_n(k) | Z^N(k-1), Z^{n-1}(k)) + \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \phi(P_e(k)) \\ &\leq \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \sum_{n=1, N} H(Z_n(k) | Z_n(k-1)) + \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \phi(P_e(k)) \\ &\stackrel{(c)}{\leq} \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \sum_{n=1, N} H(Z_n(k) | Z_n(k-1)) + \frac{1}{N} \cdot \phi \left[ \frac{1}{K} \cdot \sum_{k=1, K} P_e(k) \right] \end{aligned}$$

$$\leq H(Z_I|Z_{II}) + \frac{1}{N} \cdot \phi \left[ \frac{1}{K} \cdot \sum_{k=1, K} P_e(k) \right], \quad (8)$$

where (c) follows from the convexity of  $\phi(\cdot)$ , and where  $Z_I$  and  $Z_{II}$  are random variables with

$$\Pr\{(Z_I, Z_{II})=(z_1, z_2)\} := \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \sum_{n=1, N} \Pr\{(Z_n(k), Z_n(k-1))=(z_1, z_2)\},$$

for  $(z_1, z_2) \in \{0, 1\}^2$ . (9)

Now note that

$$\Pr\{(Z_I, Z_{II})=(1, 1)\} = 0, \text{ and that} \quad (10a)$$

$$\lim_{K \rightarrow \infty} \frac{1}{K} \Pr\{Z_I=z\} = \lim_{K \rightarrow \infty} \frac{1}{K} \Pr\{Z_{II}=z\} \text{ for } z \in \{0, 1\}. \quad (10b)$$

Therefore assume that  $\lim_{K \rightarrow \infty} \frac{1}{K} \Pr\{(Z_I, Z_{II})=(0, 1)\} = \lim_{K \rightarrow \infty} \frac{1}{K} \Pr\{(Z_I, Z_{II})=(1, 0)\} = \alpha$ , and  $\lim_{K \rightarrow \infty} \frac{1}{K} \Pr\{(Z_I, Z_{II})=(0, 0)\} = 1 - 2\alpha$ , for some  $\alpha$  with  $0 \leq \alpha \leq 1/2$ . Hence, taking limits for  $K \rightarrow \infty$ , we obtain from (8)

$$\frac{1}{N} \cdot \log(M) \leq (1-\alpha) \cdot h(\alpha/(1-\alpha)) + \frac{1}{N} \cdot \phi \left[ \lim_{K \rightarrow \infty} \frac{1}{K} \cdot \sum_{k=1, K} P_e(k) \right]. \quad (11)$$

Therefore for any T-achievable rate R for all (small enough)  $\delta > 0$  there exists an  $0 \leq \alpha \leq 1/2$  such that

$$R - \delta \leq \frac{1}{N} \cdot \log(M) \leq (1-\alpha) \cdot h(\alpha/(1-\alpha)) + \frac{1}{N} \cdot \phi(\delta), \quad (12)$$

hence we can conclude that for all T

$$C_{\text{informed, informed}}^{(T)} \leq \max_{\alpha} (1-\alpha) \cdot h(\alpha/(1-\alpha)) \stackrel{(d)}{=} \log((1+\sqrt{5})/2). \quad (13)$$

Equality (d) follows from simple analysis.

## 5. Proof of Lemma 2

For  $k = 1, 2, \dots$  we have that

$$\begin{aligned} \log(M) &= H(W(k)) = I(W(k); Z^N(k)) + H(W(k) | Z^N(k)) \\ &\stackrel{(e)}{=} I(W(k); Z^N(k)) + H(W(k) | Z^N(k), \hat{W}(k)) \\ &\stackrel{(f)}{\leq} I(W(k); Z^N(k)) + \phi(P_e(k)) \\ &= \sum_{n=1, N} I(W(k); Z_n(k) | Z^{n-1}(k)) + \phi(P_e(k)) \\ &= \sum_{n=1, N} \left[ H(Z_n(k) | Z^{n-1}(k)) - H(Z_n(k) | Z^{n-1}(k), W(k)) \right] + \phi(P_e(k)), \quad (14) \end{aligned}$$

where (e) follows from (3a), and where (f) is Fano's inequality. Next consider

$$\begin{aligned} H(Z_n(k) | Z^{n-1}(k), W(k)) &\geq H(Z_n(k) | Z^{n-1}(k), W(k), G_n(k), Z^{n-1}(k-1)) \\ &= \Pr\{G_n(k)=i\} \cdot H(Z_n(k) | Z^{n-1}(k), W(k), G_n(k)=i, Z^{n-1}(k-1)) \\ &= \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k), W(k), G_n(k)=i, Z^{n-1}(k-1)) \\ &\stackrel{(g)}{=} \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k-1)), \quad (15) \end{aligned}$$

where (g) follows from

$$\begin{aligned} &P \left[ z^{n-1}(k-1), z_n(k-1), w(k), g_n(k), z^{n-1}(k) \right] \\ &= P \left[ z^{n-1}(k-1), z_n(k-1) \right] \cdot P \left[ w(k), g_n(k) \right] \cdot P \left[ z^{n-1}(k) | z^{n-1}(k-1), w(k) \right] \\ &= P \left[ z^{n-1}(k-1) \right] \cdot P \left[ z_n(k-1) | z^{n-1}(k-1) \right] \cdot P \left[ w(k), g_n(k), z^{n-1}(k) | z^{n-1}(k-1) \right]. \quad (16) \end{aligned}$$

From (14) and (15) we conclude that

$$\begin{aligned} & \log(M) - \phi(P_e(k)) \\ & \leq \sum_{n=1, N} \left[ H(Z_n(k) | Z^{n-1}(k)) - \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k-1)) \right]. \end{aligned} \quad (17)$$

This implies that (note that (h) follows from the convexity of  $\phi(\cdot)$ )

$$\begin{aligned} & \frac{1}{N} \cdot \log(M) - \frac{1}{N} \cdot \phi\left[\frac{1}{K} \sum_{k=1, K} P_e(k)\right] \\ & \stackrel{(h)}{\leq} \frac{1}{N} \cdot \log(M) - \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \phi(P_e(k)) \\ & \leq \frac{1}{K \cdot N} \cdot \sum_{k=1, K} \sum_{n=1, N} \left[ H(Z_n(k) | Z^{n-1}(k)) - \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k-1)) \right] \\ & = \frac{1}{N} \cdot \sum_{n=1, N} \frac{1}{K} \cdot \sum_{k=1, K} \left[ H(Z_n(k) | Z^{n-1}(k)) - \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k-1)) \right] \\ & \leq \max_n \frac{1}{K} \cdot \sum_{k=1, K} \left[ H(Z_n(k) | Z^{n-1}(k)) - \Pr\{G_n(k)=i\} \cdot H(Z_n(k-1) | Z^{n-1}(k-1)) \right] \\ & \leq \max_n \frac{1}{K} \cdot \left[ \sum_{k=2, K} (1 - \Pr\{G_n(k)=i\}) \cdot H(Z_n(k-1) | Z^{n-1}(k-1)) + 1 \right] \\ & \leq \frac{1}{K} \cdot \left[ \sum_{k=2, K} (1 - \Pr\{G(k)=i\}) \cdot H(Z(k-1)) + 1 \right], \end{aligned} \quad (18)$$

for some set of random variables  $(Z(0), G(1), Z(1), \dots, Z(K-1), G(K))$  with distribution

$$\begin{aligned} P[z(0), g(1), z(1), \dots, z(K-1), g(K)] &= P[z(0)] \cdot P[g(1)] \cdot P[z(1) | z(0), g(1)] \cdot \dots \\ & \quad \cdot P[g(k-1)] \cdot P[z(K-1) | z(K-2), g(K-1)] \cdot P[g(k)]. \end{aligned} \quad (19)$$

From the fact that  $T = 1$  we obtain that there must exist a  $\beta$ ,  $0 \leq \beta \leq 1$ , such that

$$\Pr\{G(k)=i\} = \beta, \text{ for } k = 1, 2, \dots \quad (20a)$$

Note that the right side of (18) is equal to  $1/K$  when  $\beta = 1$ . Therefore we assume in what follows that  $\beta < 1$ . We will now determine  $\lim_{K \rightarrow \infty} \Pr\{Z(k)=1\}$ . Define

$$\gamma_k := \Pr\{Z(k)=1\}, \quad k = 0, 1, 2, \dots \quad (20b)$$

then

$$\begin{aligned} \gamma_0 &= 0 \text{ and} \\ \gamma_k &= \Pr\{Z(k)=1\} = \Pr\{Z(k-1)=0\} \cdot \Pr\{G(k)=i\} \\ &= (1 - \Pr\{Z(k-1)=1\}) \cdot \Pr\{G(k)=i\} \\ &= (1 - \gamma_{k-1}) \cdot \beta, \quad k = 1, 2, \dots \end{aligned} \quad (21)$$

Define  $\Gamma := \beta/(1+\beta)$  and  $\delta_k := \gamma_k - \Gamma$ . Then

$$\delta_{k+1} = \gamma_{k+1} - \Gamma = (1 - \gamma_k) \cdot \beta - (1 - \Gamma) \cdot \beta = (\Gamma - \gamma_k) \cdot \beta = -\delta_k \cdot \beta. \quad (22)$$

Therefore  $|\delta_{k+1}| < |\delta_k|$  and  $\lim_{K \rightarrow \infty} \gamma_k = \beta/(1+\beta)$ . Now taking in (18) the limit for  $K \rightarrow \infty$  at both sides we find that

$$\begin{aligned} &\frac{1}{N} \cdot \log(M) \\ &\leq \lim_{K \rightarrow \infty} \frac{1}{K} \cdot \left[ \sum_{k=2, K} (1 - \Pr\{G(k)=i\}) \cdot H(Z(k-1)) + 1 \right] + \frac{1}{N} \cdot \lim_{K \rightarrow \infty} \phi \left[ \frac{1}{K} \sum_{k=1, K} P_e(k) \right] \\ &= (1 - \beta) \cdot h(\beta/(1+\beta)) + \frac{1}{N} \cdot \phi \left[ \lim_{K \rightarrow \infty} \frac{1}{K} \sum_{k=1, K} P_e(k) \right]. \end{aligned} \quad (23)$$

Note that in the limit for  $K \rightarrow \infty$ , (23) holds for both  $\beta < 1$  and  $\beta = 1$ .

Therefore for any 1-achievable rate  $R$  for all (small enough)  $\delta > 0$  there exists a  $0 \leq \beta \leq 1$  such that

$$R - \delta \leq \frac{1}{N} \cdot \log(M) \leq (1 - \beta) \cdot h(\beta/(1+\beta)) + \frac{1}{N} \cdot \phi(\delta). \quad (24)$$

From this we can conclude that

$$C_{\text{uninformed,uninformed}}^{(T=1)} \leq \max_{\beta} (1-\beta) \cdot h(\beta/(1+\beta)). \quad (25)$$

Numerical computation shows that  $\max_{\beta} (1-\beta) \cdot h(\beta/(1+\beta)) = 0.54588\dots$  is achieved for  $\beta = 0.2887\dots$

## 6. Conclusion

We have shown that also in the average-error case  $\log_2((1+\sqrt{5})/2)$  is an upper bound for the T-capacities in all four (Simonyi-) situations. In addition we found an upper bound for the average-error 1-capacity in the case where both the encoder and the decoder are uninformed.

Wyner and Ozarow [7] independently found an upper bound for the average-error capacity in the uninformed-uninformed case. Their proof is more concise than ours, but not detailed as far as the limiting behavior (for  $K \rightarrow \infty$ ) is concerned. Berger [8] informed the author about the existence of this unpublished material.

Van Overveld [9] demonstrated that also the 2-capacity is upper-bounded by 0.54588... in the uninformed-uninformed case. It is still unknown however whether this holds for all T-capacities or not.

## 7. Acknowledgement

Part of the research for this report was performed at Bielefeld University in December 1987. Rudy Ahlswede and Zhen Zhang, thank you for inviting me, and for providing the right atmosphere.

## References

- [1] S. Baggen (Philips Research Laboratories, Eindhoven, The Netherlands), private communication (December 1985).
- [2] F.M.J. Willems & A.J. Vinck, "Repeated recording for an optical disc," **Proc. 7th Symp. on Information Theory in the Benelux**, Noordwijkerhout, The Netherlands, May 22-23, 1986. Ed. by D.E. Boeke, Delft University Press, 1986, p. 49-53.

- 
- [3] S.I. Gel'fand, "Capacity of one broadcast channel," **Problems of Information Transmission**, Vol. 13, (1977), p. 240-242. Transl. of **Problemy Peredachi Informatsii**, Vol. 13, No. 3 (July-September 1977), p. 106-108.
  - [4] M.S. Pinsker, "Capacity of noiseless broadcast channels," **Problems of Information Transmission**, Vol. 14, (1978), p. 97-102. Transl. of **Problemy Peredachi Informatsii**, Vol. 14, No. 2 (April-June 1978), p. 28-34.
  - [5] M. Borden, "Coding for write-unidirectional memories." Submitted for publication in **IEEE Trans. Inform. Theory**.
  - [6] G. Simonyi, "On write-unidirectional memory codes," preprint 35/1987, Math. Inst. of the Hungarian Acad. of Sciences, Budapest, Hungary. Submitted for publication in **IEEE Trans. Inform. Theory**.
  - [7] A.D. Wyner & L.H. Ozarow (AT&T Bell Laboratories, Murray Hill, N.J., U.S.A.), private communication (March 1988).
  - [8] T. Berger (Cornell University, Ithaca, N.Y, U.S.A.), private communication (February 1988).
  - [9] W.M.C.J. van Overveld (Information Theory Group, Faculty of Electrical Engineering, Eindhoven University of Technology, Eindhoven, The Netherlands), private communication (January 1988).

- (188) Jóźwiak, J.  
THE FULL DECOMPOSITION OF SEQUENTIAL MACHINES WITH THE STATE AND OUTPUT BEHAVIOUR REALIZATION.  
EUT Report 88-E-188. 1988. ISBN 90-6144-188-9
- (189) Pineda de Gyvez, J.  
ALWAYS: A system for wafer yield analysis.  
EUT Report 88-E-189. 1988. ISBN 90-6144-189-7
- (190) Siuzdak, J.  
OPTICAL COUPLERS FOR COHERENT OPTICAL PHASE DIVERSITY SYSTEMS.  
EUT Report 88-E-190. 1988. ISBN 90-6144-190-0
- (191) Bastiaans, M.J.  
LOCAL-FREQUENCY DESCRIPTION OF OPTICAL SIGNALS AND SYSTEMS.  
EUT Report 88-E-191. 1988. ISBN 90-6144-191-9
- (192) Worm, S.C.J.  
A MULTI-FREQUENCY ANTENNA SYSTEM FOR PROPAGATION EXPERIMENTS WITH THE OLYMPUS SATELLITE.  
EUT Report 88-E-192. 1988. ISBN 90-6144-192-7
- (193) Kersten, W.F.J. and C.A.P. Jacobs  
ANALOG AND DIGITAL SIMULATION OF LINE-ENERGIZING OVERVOLTAGES AND COMPARISON WITH MEASUREMENTS IN A 400 kV NETWORK.  
EUT Report 88-E-193. 1988. ISBN 90-6144-193-5
- (194) Hosselet, L.M.L.F.  
MARTINUS VAN MARUM: A Dutch scientist in a revolutionary time.  
EUT Report 88-E-194. 1988. ISBN 90-6144-194-3
- (195) Bondarev, V.N.  
ON SYSTEM IDENTIFICATION USING PULSE-FREQUENCY MODULATED SIGNALS.  
EUT Report 88-E-195. 1988. ISBN 90-6144-195-1
- (196) Liu Wen-Jiang, Zhu Yu-Cai and Cai Da-Wei  
MODEL BUILDING FOR AN INGOT HEATING PROCESS: Physical modelling approach and identification approach.  
EUT Report 88-E-196. 1988. ISBN 90-6144-196-X
- (197) Liu Wen-Jiang and Ye Dau-Hua  
A NEW METHOD FOR DYNAMIC HUNTING EXTREMUM CONTROL, BASED ON COMPARISON OF MEASURED AND ESTIMATED VALUE.  
EUT Report 88-E-197. 1988. ISBN 90-6144-197-8
- (198) Liu Wen-Jiang  
AN EXTREMUM HUNTING METHOD USING PSEUDO RANDOM BINARY SIGNAL.  
EUT Report 88-E-198. 1988. ISBN 90-6144-198-6
- (199) Jóźwiak, L.  
THE FULL DECOMPOSITION OF SEQUENTIAL MACHINES WITH THE OUTPUT BEHAVIOUR REALIZATION.  
EUT Report 88-E-199. 1988. ISBN 90-6144-199-4
- (200) Huis in 't Veld, R.J.  
A FORMALISM TO DESCRIBE CONCURRENT NON-DETERMINISTIC SYSTEMS AND AN APPLICATION OF IT BY ANALYSING SYSTEMS FOR DANGER OF DEADLOCK.  
EUT Report 88-E-200. 1988. ISBN 90-6144-200-1
- (201) Woudenberg, H. van and R. van den Born  
HARDWARE SYNTHESIS WITH THE AID OF DYNAMIC PROGRAMMING.  
EUT Report 88-E-201. 1988. ISBN 90-6144-201-X
- (202) Engelshoven, R.J. van and R. van den Born  
COST CALCULATION FOR INCREMENTAL HARDWARE SYNTHESIS.  
EUT Report 88-E-202. 1988. ISBN 90-6144-202-8
- (203) Delissen, J.G.M.  
THE LINEAR REGRESSION MODEL: Model structure selection and biased estimators.  
EUT Report 88-E-203. 1988. ISBN 90-6144-203-6
- (204) Kalasek, V.K.I.  
COMPARISON OF AN ANALYTICAL STUDY AND EMTF IMPLEMENTATION OF COMPLICATED THREE-PHASE SCHEMES FOR REACTOR INTERRUPTION.  
EUT Report 88-E-204. 1988. ISBN 90-6144-204-4

- (205) Butterweck, H.J. and J.H.F. Ritzerfeld, M.J. Werter  
FINITE WORDLENGTH EFFECTS IN DIGITAL FILTERS: A review.  
EUT Report 88-E-205. 1988. ISBN 90-6144-205-2
- (206) Bollen, M.H.J. and G.A.P. Jacobs  
EXTENSIVE TESTING OF AN ALGORITHM FOR TRAVELLING-WAVE-BASED DIRECTIONAL  
DETECTION AND PHASE-SELECTION BY USING TWONFIL AND EMTF.  
EUT Report 88-E-206. 1988. ISBN 90-6144-206-0
- (207) Schuurman, W. and M.P.H. Weenink  
STABILITY OF A TAYLOR-RELAXED CYLINDRICAL PLASMA SEPARATED FROM THE WALL  
BY A VACUUM LAYER.  
EUT Report 88-E-207. 1988. ISBN 90-6144-207-9
- (208) Lucassen, F.H.R. and H.H. van de Ven  
A NOTATION CONVENTION IN RIGID ROBOT MODELLING.  
EUT Report 88-E-208. 1988. ISBN 90-6144-208-7
- (209) Jóźwiak, L.  
MINIMAL REALIZATION OF SEQUENTIAL MACHINES: The method of maximal  
adjacencies.  
EUT Report 88-E-209. 1988. ISBN 90-6144-209-5
- (210) Lucassen, F.H.R. and H.H. van de Ven  
OPTIMAL BODY FIXED COORDINATE SYSTEMS IN NEWTON/EULER MODELLING.  
EUT Report 88-E-210. 1988. ISBN 90-6144-210-9
- (211) Boom, A.J.J. van den  
 $H_{\infty}$ -CONTROL: An exploratory study.  
EUT Report 88-E-211. 1988. ISBN 90-6144-211-7
- (212) Zhu Yu-Cai  
ON THE ROBUST STABILITY OF MIMO LINEAR FEEDBACK SYSTEMS.  
EUT Report 88-E-212. 1988. ISBN 90-6144-212-5
- (213) Zhu Yu-Cai, M.H. Driessen, A.A.H. Damen and P. Eykhoff  
A NEW SCHEME FOR IDENTIFICATION AND CONTROL.  
EUT Report 88-E-213. 1988. ISBN 90-6144-213-3
- (214) Bollen, M.H.J. and G.A.P. Jacobs  
IMPLEMENTATION OF AN ALGORITHM FOR TRAVELLING-WAVE-BASED DIRECTIONAL  
DETECTION.  
EUT Report 89-E-214. 1989. ISBN 90-6144-214-1
- (215) Hoeijmakers, M.J. en J.M. Vleeshouwers  
EEN MODEL VAN DE SYNCHRONE MACHTINE MET GELIJKRICHTER, GESCHIKT VOOR  
REGELEDOELEINDEN.  
EUT Report 89-E-215. 1989. ISBN 90-6144-215-X
- (216) Pineda de Gyvez, J.  
LASER: A LAYout Sensitivity ExploreR. Report and user's manual.  
EUT Report 89-E-216. 1989. ISBN 90-6144-216-8
- (217) Duarte, J.L.  
MINAS: An algorithm for systematic state assignment of sequential  
machines - computational aspects and results.  
EUT Report 89-E-217. 1989. ISBN 90-6144-217-6
- (218) Kamp, M.M.J.L. van de  
SOFTWARE SET-UP FOR DATA PROCESSING OF DEPOLARIZATION DUE TO RAIN  
AND ICE CRYSTALS IN THE OLYMPUS PROJECT.  
EUT Report 89-E-218. 1989. ISBN 90-6144-218-4
- (219) Koster, G.J.P. and L. Stok  
FROM NETWORK TO ARTWORK: Automatic schematic diagram generation.  
EUT Report 89-E-219. 1989. ISBN 90-6144-219-2
- (220) Willems, F.M.J.  
CONVERSEES FOR WRITE-UNIDIRECTIONAL MEMORIES.  
EUT Report 89-E-220. 1989. ISBN 90-6144-220-6