Fast and robust method for measuring semiconductor optical amplifier gain

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Fast and Robust Method for Measuring Semiconductor Optical Amplifier Gain

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Abstract—In this paper, we present a new, robust multipoint fitting method for gain measurement with a metric for quality estimation of the procedure. The method is able to identify the deleterious effect of imperfections within the test structures, is tolerant to optical coupling errors and is well suited to high throughput, generic, automated testing of semiconductor optical amplifiers. Gain is estimated in a range of pump current densities over multiple spectral bands from 1400 to 1600 nm with a standard error in the order of 1/cm.

Index Terms—Gain measurement, semiconductor optical amplifiers, photonic integrated circuits, segmented device, active-passive integration.

I. INTRODUCTION

SEMICONDUCTOR optical amplifiers (SOAs) play an important role as a component in photonic integrated circuits. They serve as a light source and as an amplifier and form the gain section of various lasers [1]. Fast and reliable methods for measuring SOA parameters, in particular, its gain, are needed to characterize SOA performance for accurate circuit design and process optimization.

Several methods to measure gain versus wavelength of integrated semiconductor optical amplifiers have been proposed. They include transmission-based methods [2], the Hakki–Paoli method and spectral-fringe-based methods [3]–[9] based on analysis of an amplified spontaneous emission (ASE) spectrum in a Fabry–Pérot (FP) cavity. A number of additional methods are based on the comparison of different SOA lengths [10]–[15]. These methods rely on assumptions about the facet quality, variations between structures and optical coupling and do not provide any information about the quality of the measurement. Moreover, they can require using a high resolution optical spectrum analyser, precision alignment and in the case of variable length SOAs, repeated optical coupling.

In this paper, we present a fast and robust method for SOA gain measurement based on the multi-point fitting of measurements from a single multi-section device (Fig. 1). The measurement requires only one optical coupling, and provides a metric for quality control of the measurement procedure. Using the proposed method, we show that measurement errors can be identified when using test structures with absorbing electrical isolation, motivating the use of optically passive waveguides between amplifiers sections.

A fabricated chip with test structures mounted on the measurement setup is shown in Fig. 1. The test cell shows an example multi-section SOA at the top of the chip with four electrical probes and just one optical connection at the left facet. Through an appropriate combination of measurements, the gain may be identified as a function of wavelength and current, with an indication of the quality of fit.

This paper is organized as follows. In Section II we give an overview of the current methods for SOA gain measurement. This is followed in Section III by a detailed description of variable SOA length methods for evaluating gain, including the structures proposed to implement it and the ways to calculate gain from the measured ASE spectra. In Section IV we study the sensitivity to coupling loss variations of different test structures, which implement a set of SOAs of different length, and identify those with tolerance to the coupling efficiency variation. Section V covers details of the test structure design and fabrication. Sections VI and VII describe the measurements we have performed and their results. The paper is completed with conclusions in Section VIII.

II. METHODS OVERVIEW

The most intuitive approach to measure SOA gain is to scan a single wavelength signal and to measure the difference between input and output levels. More sophisticated, high-throughput
transmission methods [2] measure gain directly over a large spectral range with spectrally broad laser probe light and require use of femtosecond pulsed laser. The measurement however combines unknown coupling losses and gain from the amplifier.

For amplifiers in Fabry–Pérot cavities several other methods can be used. The threshold current for a set of devices of different cavity length can be evaluated [16]. If the total optical loss in the cavity is known for each length, the plot of optical loss versus threshold current gives a relation of peak gain and current density.

A group of methods based on the analysis of the amplified spontaneous emission (ASE) spectrum in a Fabry–Pérot (FP) cavity has been studied intensively in the literature. In the method proposed by Hakki and Paoli [3], a spectral gain distribution is obtained from the peak-to-valley ratio of each cavity mode of the ASE spectrum. The measured peak height is sensitive to the resolution of the optical spectrum analyzer and therefore the Hakki–Paoli method may underestimate the gain value at current densities approaching the threshold current [8]. Cassidy et al. [4] have proposed using the ratio of the integral (mode sum) of the wavelength resolved power over one mode to the minimum power (instead of the peak-to-valley ratio in the Hakki–Paoli method) to derive the gain from the ASE spectrum. This method is much less sensitive to the response function of the optical spectrum analyzer (OSA) than the Hakki–Paoli method.

The Hakki–Paoli and Cassidy methods are sensitive to the noise as a result of the small intensity at the ASE spectrum minima. An improvement to reduce the sensitivity to the noise was made by Wang [5], who fitted the profile of each of the FP modes to an Airy function, thus obtaining a more accurate power value in the minimum. A method based on least square fitting was proposed by Zhang [6] to eliminate the error due to the asymmetric shape of longitudinal modes in a broader wavelength range. The noise limit problem is also addressed in the Fourier transform method and the Fourier series expansion (FSE) method [7]–[9], which calculate the gain based on the ratio of the intensity of the wavenumber-dependent DC- and first order AC-component in the transformed domain. They have a reduced noise sensitivity, as they use weighted integrals over the mode and not a single point measurement so that the noise is partly reduced by summation.

Methods for gain measurement for single electrode amplifiers have so far been sensitive either to coupling variations, or measurement equipment resolution and noise floor limit due to the small light intensity at the minimum. This has lead researchers to study multiple electrode methods. They make use of the relation between output ASE spectrum intensity and the SOA length, as proposed by Oster in the variable stripe length method [10], [11]. Thomson et al. [12] have shown that for a set of two spectra for SOAs of different lengths, the modal gain values can be obtained at every wavelength. The drawback of these variable amplifier length methods is that a time-consuming and precision-optical alignment is needed for measuring ASE from each of the stripes. To guarantee a good accuracy of the methods it is also critical to have same coupling loss for all the structures, which is hard to control due to variations in the fabrication of optical waveguides. This can be solved at the expense of space and measurement time by increasing the number of stripes used and fitting the obtained result.

As an alternative to the set of devices of different length, sectioned devices with independent control of injection were proposed both for gain and absorption measurements [13]–[15]. This approach was further developed by integrating the light source and integrated spectrometer [17] to allow all-electrical gain measurement on wafer scale.

### III. Variable Amplifier Length Methods

#### A. Amplified Spontaneous Emission Versus amplifier Length

The increase of the light intensity $P$ in an SOA is determined by the gain $g$ and the output spontaneous emission $P_{sp}$ of the amplifier (both per unit length) [18]:

$$\frac{dP}{dx} = gP + p_{sp},$$

where $P$ is light intensity in a given point within an amplifier. Gain here is the net modal gain. Provided we have no incoming light ($P(x = 0) = P_{in} = 0$), integrating (1) for an SOA of length $L$, we calculate the measured single-pass amplified spontaneous emission intensity $P_{ASE,L}$ as [10]

$$P_{ASE,L}(\lambda) = \eta_c \frac{p_{sp}(\lambda)}{g(\lambda)} (e^{g(\lambda)L} - 1),$$

where $\eta_c$ takes into account the output power coupling efficiency from the waveguide to the measurement equipment. Here, all the quantities have a wavelength dependency, therefore from (2), gain spectrum $g(\lambda)$ can be calculated from measured $P_{ASE,L}(\lambda)$ from SOAs of different length.

#### B. Test Structures

Two types of structures were used in the literature to realize a set of SOAs with different length: variable stripe length and segmented devices. Fig. 2 shows the difference between measurement procedure of these two types. The variable stripe length method [10], [11] employs multiple structures, each having a different optical output to be coupled to the measurement.

![Variable Amplifier Length Methods](image-url)
system. The measurement procedure consists of consecutive alignment to structures $a$, $b$, $c$, etc., and measurement of ASE spectra emitted by each of them (Fig. 2(a)).

A segmented contact structure [13]–[15] represents the second type of structures. It contains electrically isolated amplifier segments, which can be independently driven with an electrical current. During the measurement, light is always collected from the same output waveguide, and the effective amplifier length is varied by applying different combinations of pump current (Fig. 2(b)).

C. Gain Calculation

The gain spectrum can be calculated from a set of ASE spectra in several different ways. In this section we list the approaches which are discussed in the literature and will be compared in the present paper. Note, that the calculation procedure can be applied to the measurement results from both types of the structures described in Section III-B. This means that the same formulas can be applied to ASE spectra obtained from both variable-stripe-type or segmented-contact structures.

We consider two methods outlined by Thomson and Xin and our new multipoint fitting method.

1) Two-Point Thomson Method: Having two SOAs $a$ and $b$, one being twice as long as the other $L_a = L$ and $L_b = 2L$, the gain can be calculated directly by using Thomson’s approach [12]:

$$g(\lambda) = \frac{1}{L} \ln \left( \frac{P_{ASE,2L}(\lambda)}{P_{ASE,L}(\lambda)} - 1 \right).$$

(3)

2) Three-Point Xin Method: In the above approach any unguided spontaneous emission is not taken into account. Among the reasons of unguided emission could be the leakage current or emission into unguided waveguide modes [19]. The presence of unguided emission can lead to a systematic error in the gain calculation, and it should be taken care of at low pump current densities and gain values [19]. During measurements, the unguided radiation can be filtered out spatially when collecting light from the waveguide. A method proposed in [15] considers the unguided spontaneous emission radiation intensity $P_{unguided}$ as an addition to the amplified spontaneous emission from (2):

$$P_L = P_{ASE,L} + P_{unguided}.$$  

(4)

The unguided radiation can be eliminated in the gain calculation by comparing ASE spectra from SOA of three different lengths. SOA lengths should satisfy the following conditions: $L_a = L_1$, $L_b = L_1 + L_2$, and $L_c = L_1 + 2L_2$. The gain is then calculated as

$$g = \frac{1}{L} \ln \left( \frac{P_{L_1+2L_2} - P_{L_1}}{P_{L_1+L_2} - P_{L_1}} - 1 \right).$$

(5)

The disadvantage of above-mentioned methods is that if any of the measured values $P_L$ contains an error, it will propagate to the calculated gain value and there is no means to discover the error.

3) Multipoint Fitting Method With Quality of Fit: We propose to extract the gain by using more available SOA lengths $L_0, L_0, \ldots$ and corresponding ASE values $P_{ASE,L}$, and fitting them to either (2) or (4) by executing the following steps:

1) At each wavelength $\lambda$ we have a set of power densities $P_L$ measured from the SOAs of length $L_i$.
2) Calculate the measurement error for each of the data points. The error is determined by the limit of the OSA sensitivity $P_{noise}$ and measurement uncertainty $\epsilon$: $\sigma_i = P_{noise} + \epsilon P_L$.
3) Calculate weight for each data point, which is inverse of the total measurement error: $w_i = 1/\sigma_i$.
4) Perform weighted non-linear least-square fit of the data points, using (2) or (4) as a model function.
5) As a result of fitting, for every fit parameter we obtain its magnitude and standard error.

The steps described above are repeated for every wavelength. The fitting approach has an intrinsic indication of the fitting quality, namely the standard error, which is calculated based on the residual sum of squares. It can be used to discard poor measurement results, e.g. due to a damaged waveguide in the middle of test structure or varying light collection efficiency.

In our paper we compare three mentioned methods: Thomson’s (3), Xin’s (5), and fitting to (2) and (4).

IV. SENSITIVITY TO COUPLING VARIATION

The calculation methods described in III-C commonly assume that the coupling efficiency $\eta_c$ is the same for each measurement condition. The use of one output waveguide removes the need for structure-to-structure realignments, but the coupling efficiency is still affected for example by the chip displacement due to the height variations in a thermoelectric cooler for a temperature stabilised measurement. In this work we perform active realignment to the best coupling position after every pump current setting.

In the following section we perform a simulation to discover dependency of the test structure design, in particular, of SOA section lengths and their number, on the optical alignment induced error.

A. Simulation Procedure

In order to estimate the effect of coupling variation on the calculated gain, we performed a Monte-Carlo simulation of the measurement procedure, while varying the fiber location around optimum.

In the simulation we vary the relative offset of the fiber with respect to the center of the waveguide in the transverse and lateral directions. These two offsets are assumed to be independent and have a normal distribution with the mean value of zero. The random offset value gives a random variation of the coupling efficiency $\eta_c$ in (2). The standard deviation for the offsets is selected so that the coupling efficiency varies by less than 0.1 dB.
Fig. 3. Monte-Carlo simulation of the calculated gain distribution for two different designs of multi-section structures.

TABLE I

<table>
<thead>
<tr>
<th># SOA lengths</th>
<th>Maximum length, ( L_{\text{max}} )</th>
<th>SOA lengths, ( L_i )</th>
</tr>
</thead>
<tbody>
<tr>
<td>2</td>
<td>400</td>
<td>200, 400</td>
</tr>
<tr>
<td>2</td>
<td>600</td>
<td>300, 600</td>
</tr>
<tr>
<td>2</td>
<td>800</td>
<td>400, 800</td>
</tr>
<tr>
<td>3</td>
<td>300</td>
<td>100, 200, 300</td>
</tr>
<tr>
<td>3</td>
<td>450</td>
<td>150, 300, 450</td>
</tr>
<tr>
<td>3</td>
<td>600</td>
<td>200, 400, 600</td>
</tr>
<tr>
<td>3</td>
<td>800</td>
<td>266, 533, 800</td>
</tr>
<tr>
<td>3</td>
<td>400</td>
<td>100, 200, 400</td>
</tr>
<tr>
<td>3</td>
<td>600</td>
<td>150, 300, 600</td>
</tr>
<tr>
<td>3</td>
<td>800</td>
<td>200, 400, 800</td>
</tr>
<tr>
<td>4</td>
<td>400</td>
<td>100, 200, 300, 400</td>
</tr>
<tr>
<td>4</td>
<td>600</td>
<td>150, 300, 450, 600</td>
</tr>
<tr>
<td>4</td>
<td>800</td>
<td>200, 400, 600, 800</td>
</tr>
<tr>
<td>6</td>
<td>600</td>
<td>100, 200, 300, ..., 600</td>
</tr>
<tr>
<td>6</td>
<td>800</td>
<td>133, 266, 400, ..., 800</td>
</tr>
<tr>
<td>10</td>
<td>800</td>
<td>700, 710, ..., 790, 800</td>
</tr>
</tbody>
</table>

All lengths are in \( \mu \text{m} \).

The simulated ASE power values \( P_{\text{ASE}, L_i} \) are calculated at a single wavelength from (2) with typical nominal values of gain \( g_n \), spontaneous emission \( p_{sp,n} \), and a set of lengths \( L_a, L_b, \ldots \). From a set of ASE power values we calculate the gain \( g_n \) and spontaneous emission \( p_{sp,s} \) using our fitting method described in the Section III-C3.

The deviations of the simulated values from the nominal values are represented as relative errors \( \delta g = (g_n - g_s)/g_n \) and \( \delta p_{sp} = (p_{sp,n} - p_{sp,s})/p_{sp,n} \).

An example of the Monte-Carlo simulation result with 5000 trials is shown in Fig. 3. From this sample we plot a histogram and calculate the standard deviation \( \sigma_g \) of the gain from the nominal value, which is an indicator of gain measurement precision for a specified misalignment.

B. Simulation Results

We carried out the Monte-Carlo simulation for different sets of SOA lengths varying the number of lengths and the maximum total length (Table I). In the table, the number of SOA lengths determines the number of data points \( P_{\text{ASE}, L_i} \) used to calculate the gain.

The corresponding standard deviation for the gain is plotted in Fig. 4 as a function of the maximum length \( L_{\text{max}} \). The standard deviation is a direct measure of the sensitivity of the measurement technique to alignment variations. Fig. 4 shows that test structures with relatively short sections lead to a higher spread in gain variation and therefore to a reduced accuracy. The structures with larger maximum length have a lower standard deviation of the gain. For a given maximum length, a clear improvement in precision is observed as the number of sections is increased.

However there is a number of factors that limit the maximum length and the number of sections. The maximum length in our case was limited by the lasing starting in the cavity formed by waveguide facets at a higher pump currents. Heat generated in the SOAs may also play a negative role by changing the gain profile of the amplifiers. ASE-induced gain saturation should also be considered and can be reduced when selecting the maximum section lengths [20]. A higher number of sections requires more electrical probes to be used in the measurement. For these reasons devices with 4 and 5 sections are selected for this study.

V. DESIGN AND FABRICATION

A. Structure Design

In our experiment we used the multi-section devices, consisting of a single waveguide with several SOA sections grown by active-passive integration. The schematic of the structures is shown in Fig. 5. The thin blue line denotes a waveguide, the red stripes denote the parts of the waveguide within an active region, and gray boxes represent the SOA electrical contact pads. The SOA sections are separated by the electrical isolation regions, which provide the gap between the SOA electrodes of 30 \( \mu \text{m} \) and ensure electrical isolation between the SOA sections. This allows them to be pumped independently. The left side output
Fig. 6. Test structure mask designs with (a) passive and (b) active isolation sections. SOA lengths in μm: (a)—100, 100, 100, 100; (b)—200, 200, 400, 200, 200.

In the paper we compare two structures. The first structure is depicted in Fig. 5(a). The absorption in the passive isolation sections between the amplifiers is small and can be neglected. The second structure is shown in Fig. 5(b). The isolation regions are made with the same active material as the amplifiers. The mask designs of the actual structures is shown in Fig. 6. Here, the black areas on top of the waveguides represent the active material location. First structure (Fig. 6(a)) contains 4 sections each having same length of 100 μm. The total SOA lengths \( L_i \) achievable with it are therefore 100, 200, 300, and 400 μm. The second, all-active, structure (Fig. 6(b)) contains sections with lengths of 200, 200, 400, 200, and 200 μm.

B. Fabrication

InP-based test structures were fabricated by Smart Photonics in a multi-project wafer (MPW) run through the JePPIX.eu service [21]. The foundry provides active-passive integrated circuit, which allows lengths of active sections to be varied independently of the chip size. The structure design is realized with standard SOAs and isolation building blocks available from the design kit provided by the foundry.

VI. MEASUREMENT SETUP

In the measurement setup (Fig. 7), the chip is epoxy-bonded to a copper chuck to ensure electrical and thermal contact. In order to keep the chip temperature constant at different current densities, an active cooling by a thermo-electrical cooler (TEC) and a thermocontroller with a Thorlabs Pro8000 Mainframe is used. Electrical signals from the source meters are provided through the individual electrical probes contacting the metal pads of the SOA sections.

A photograph of the chip installed on the measurement setup with the probe needles touching SOA contact pads is shown in Fig. 1.

Light emitted by the SOAs is collected through the chip facet with a single-mode lensed optical fiber. The fiber is mounted on the 3-axis piezo positioner, which allows accurate fiber alignment to the waveguide. An SOA emits in both TE and TM polarizations. We have measured the ratio of TE/TM amplified spontaneous emission intensity for a range of SOA lengths, which shows more than 10 dB difference between TE and TM gain and spontaneous emission over the 1500–1600 nm range. This translates into a systematic error not exceeding 5%. A polarization filter may be used to separately measure TE and TM gain. To accommodate the temperature expansion of the mount and to keep coupling efficiency from the integrated waveguide to the lensed fiber constant, the fiber tip is realigned after each current setting. The collected light is then directed to an optical spectrum analyser ANDO AQ-6315A, which measures the resulting ASE spectrum with a resolution of 0.5 nm. The OSA sensitivity was measured to be \( P_{noise} = -90 \) dBm and its measurement uncertainty is \( \epsilon = 0.3 \) dB from the specification.

VII. MEASUREMENT RESULTS

A. ASE Spectra Measurement

ASE is generated by the SOAs for a range of pump current densities \( J \). In our experiments we covered the range from 0.25 to 10 kA/cm².

Currents are applied to SOA sections in 4 different configurations of electrode bias (Table II). In the table, the sections are listed starting from the optical output side. The sections are either forward biased with a current \( I \) or grounded (0 V). An example of measured ASE spectra from the structure in Fig. 5(a) at a current density \( J = 10 \) kA/cm² is shown in Fig. 8. The spectra are for four different SOA lengths. This ensures four data points per wavelength for further gain calculation.

B. Gain Calculation

To compare our fitting method with other methods, we calculate the gain at each wavelength using four different approaches:
Fig. 8. Measured ASE for gain calculation for four different lengths. Pump current density is $J = 10$ kA/cm$^2$ in each case.

i) Two-point Thompson method (3) from configurations $a$ and $b$. SOA lengths are $L = L_a = 100$ $\mu$m, and $2L = L_b = 200$ $\mu$m;

ii) Two-point Thompson method (3) from configurations $b$ and $d$. SOA lengths are $L = L_b = 200$ $\mu$m, and $2L = L_d = 400$ $\mu$m;

iii) Three-point Xin method (5) from configurations $a$, $b$, and $c$. SOA lengths are $L_1 = L_a = 100$ $\mu$m, $L_1 + L_2 = L_b = 200$ $\mu$m, and $L_1 + 2L_2 = L_c = 300$ $\mu$m;

iv) our method, four-point fitting to (2) from all configurations $a$ to $d$ are used. We followed the procedure described in the Section III-C3. An open-source Python package [22] was used to perform least-square fitting and calculate the standard error (SE).

Gain spectra have been derived from the ASE measurement data using the four methods as specified above, and these are compared in Fig. 9(a) for three different bias current densities. Net modal gain is defined as the optical gain $g$ in (1). On first inspection, the agreement over the 3 dB spectral width between the estimates is good, particularly for the 4.5 kA/cm$^2$ bias point. However there are significant differences between the methods outside of the gain peak and at low and high values of current. Furthermore there is limited indication of the quality of the data.

We have no $a$-priori way to know which method gives the correct result. A commonly used method for measurement in the wide range of current densities is the two-point Thomson method, which we use as our reference method. However there are two length pairs ($L = 100, 200$ $\mu$m and $L = 200, 400$ $\mu$m) to which this method can be applied, since it is not known in advance which pair is best, our reference gain value is their mean value. The deviation of the gain calculated with the four methods from this reference value for the selected current density $J = 4.5$ kA/cm$^2$ is plotted as gain difference in Fig. 9(b). All four methods show a gain difference within $\pm 1$ cm$^{-1}$ within the top part of gain spectrum. The difference from the reference increases towards the edges of the wavelength range and its maximum value is within $\sim 7$ cm$^{-1}$.

Fig. 9(b) clearly shows the disagreement for the two-point Thompson measurements for the two different lengths pairs indicating a source of error even within the same test structure. Our multi-point fitting method gives values close to the reference value.

Fig. 9. Calculated net modal gain. (a) Comparison of gain calculated with different methods. (b) Difference of measured gain values from reference. Pump current density $J = 4.5$ kA/cm$^2$. (c) Gain standard error obtained from the four-point fitting method.

We compare the maximum gain differences from the reference for a range of pump current densities in Fig. 10. For this figure, the maximum gain difference was specified over wavelength ranges with gain values above $-20$ cm$^{-1}$. The differences are particularly high for current densities below 2 kA/cm$^2$, where spontaneous emission intensity and gain are low and ASE spectra reach the noise floor level of the OSA. The calculated gain therefore has a large spread, particularly at lower and higher
wavelengths, similar to the variations shown in Fig. 9(a) above 1600 nm. Fig. 10 shows that the four-point fitting method gives a much smaller variation from the reference value than two- and three-point analytical methods for the whole range of current densities.

C. Quality of Fit

Besides gain value, our fitting method allows confidence intervals to be calculated. Fig. 9(c) shows the standard error of the fitted gain, which is an important additional parameter obtained from the method. It is calculated automatically based on the residuals for each data point, and can be used for quality control of the method. This metric allows us to identify deviations from the amplified spontaneous emission model assumed in Section III-A. In case the structures are defective (due to design and/or fabrication), or, for example, the coupling was not constant during the measurements, this will show up through increased standard error values. Deviations from the approximation made in (1), such as the onset of laser oscillation or gain saturation will also be detected through this metric.

As a demonstration of this capability, we compare the two test structures described in Section V-A, one with optically passive isolation sections, and one with active isolation sections which may be expected to act as absorbing regions.

The data for the first structure with passive isolations are shown in the Fig. 9(a) and provide a control experiment. The derived gain spectra from the all-active with active isolation sections are shown in Fig. 11(a) for a current density \( J = 4.5 \) kA/cm\(^2\). Comparison of Figs. 11(a) and 9(a) shows that absorption at lower wavelengths leads to considerable difference in predicted gain for nominally the same amplifier building block design.

The standard error for the fitted gain for the second structure is shown in Fig. 11(b). It has a small value in the wavelength range above 1580 nm, where the gain measurement is consistent with data in Fig. 9(a). However the standard error increased from \( \sim 1 \) cm\(^{-1}\) in Fig. 9(c) to \( \sim 5 \) cm\(^{-1}\) in Fig. 11(b), indicating unreliable data for the range 1420 nm to 1550 nm for the all-active structure. Absorption in the active isolation sections is expected to be insignificant at longer wavelengths, but can be expected to increase at lower wavelengths. The standard error metric indeed points out the problem in the measurement. By setting a threshold value for this metric, it is possible to include the quality control in the measurement procedure, which is important for automated testing.

D. Measurement Automation

To measure the SOA net modal gain in the wide range of current densities, the described above measurements can be automated. We have applied the multi-point fitting method on the multi-section SOA test structure from Fig. 6(a) for current densities from 0.25 to 10 kA/cm\(^2\). The automated measurement procedure includes repetitive pump current setting with the particular configuration, fiber realignment, and spectrum measurement with OSA, followed by the gain calculation with a procedure described in Section III-C3. Fig. 12 shows the net modal gain obtained with the described procedure. The graph shows the expected blue shift of the peak gain value with the increased pump current density. The corresponding absolute standard error for the calculated gain is shown in Fig. 13. This parameter is below 1 cm\(^{-1}\) for the regions with positive gain, and increases towards the edges of the spectrum, particularly at lower current densities. This is due to the low intensity of spontaneous emission and low gain/absorption in the active material, which in total give the weak ASE intensity comparable to the sensitivity level of the OSA. Gain measurement at the extreme wavelengths can be improved by increasing the sensitivity of the OSA at the expense of the measurement.
time. This gives a high degree of confidence in the measurement data over the whole operating range of the SOA building block.

VIII. CONCLUSION

We presented a fitting method for calculating SOA gain based on multi-section devices. The method is more robust compared to other methods based on variable-length structures. We have shown that it provides a metric for quality estimation of the performed calculation, which is important for automation of the gain measurement. The technique is expected to be more resilient to optical alignment errors, and sensitive to imperfections such as undesired absorption in the test structure. The suitability for automation offers an attractive method for robust, high-throughput characterisation of optical amplifier building blocks for statistically more accurate data for design models and process optimisation.

REFERENCES

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